

Automotive Power Electronics Roadmap

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 Summary authored by S. D. Round, ETH Zurich, Switzerland

Automotive applications for power electronics is increasing rapidly due to the demand for hybrid and future fuel-cell powered vehicles. The power electronic systems are not only required for driving the vehicle (Fig. 1) but are also used to interface energy storage components and to supply high power auxiliary systems such as active suspension, electric valves and air conditioning units. The automotive industry has specific requirements for its power electronic systems such as a compact design, high reliability, long life time and an extremely low cost to power ratio. The systems are further required to operate over a wide ambient temperature range and with liquid cooling temperatures of typically 105°C. In a study from the USA FreedomCAR project, it is projected that the required cost of the power electronic systems has to reduce by a factor of three until the year 2020.

The task of the Automotive Roadmap Committee was to clarify which technologies are needed to achieve the performance and cost targets of the automotive industry. The road mapping effort focused on three systems as circled in Fig. 1:

1. a non-isolated dc-dc converter, in the 40 to 100 kW power range, that can be used as a fuel cell interface,
2. an ac-dc inverter that is integrated into the machine housing of a hybrid drive system (since an

integrated solution provides the greatest cost reduction potential), and,

3. an isolated dc-dc converter to provide bi-directional power flow between the high voltage bus and the 14 V accessory power system, where the required power range is 1 to 3 kW.

The main outcomes of the road mapping exercise are that the drive inverter cost target could potentially be met if the power electronics is integrated, and that the maximum achievable power density of the non-isolated dc-dc converter and the isolated dc-dc converter is 50 kW/liter and 10 kW/liter respectively.

The road mapping process utilized a bottom-up approach. Here, mathematical descriptions for the electrical, thermal, packaging and magnetic components are developed. Using these descriptions a component technology space is formed. By using the specifications, topologies, and operating parameters the component space can be optimally mapped into a system performance space, which gives system performance measures such as efficiency, power density and costs. Exploring the performance space and demanding an improved system performance, and then undertaking a reverse mapping from this new point back into the component space, provides information on how the technologies must be developed to achieve the new desired system performance.

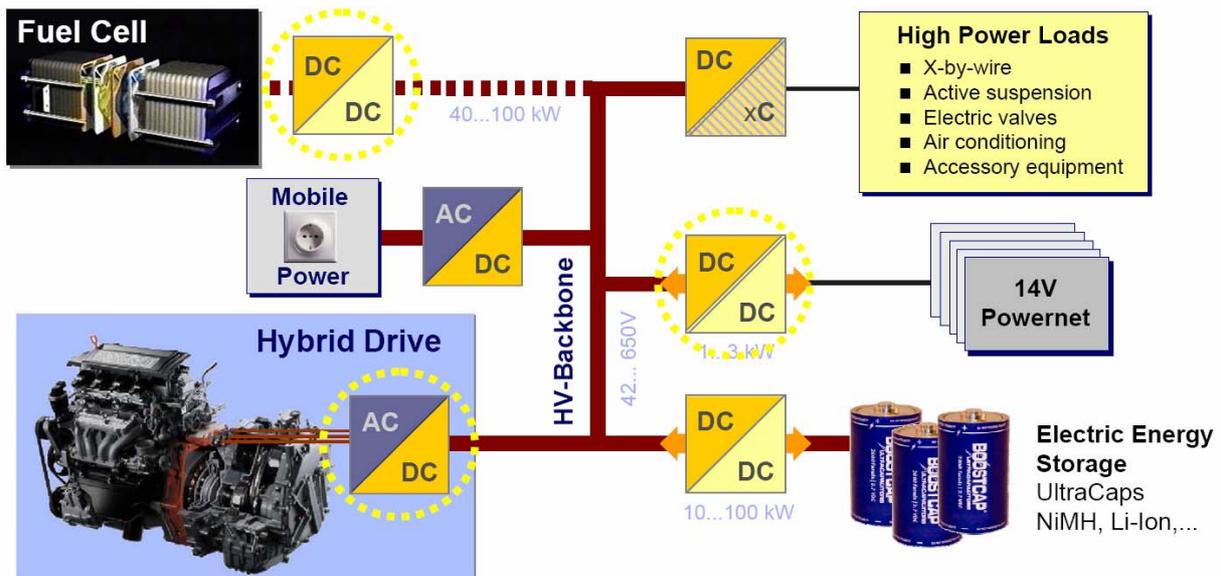


Fig. 1 Power electronic key systems for the cars of tomorrow. The three considered systems in the automotive power electronics road mapping exercise are encircled in yellow.



Automotive Power Electronics Research Roadmap Initiative

Coordinators

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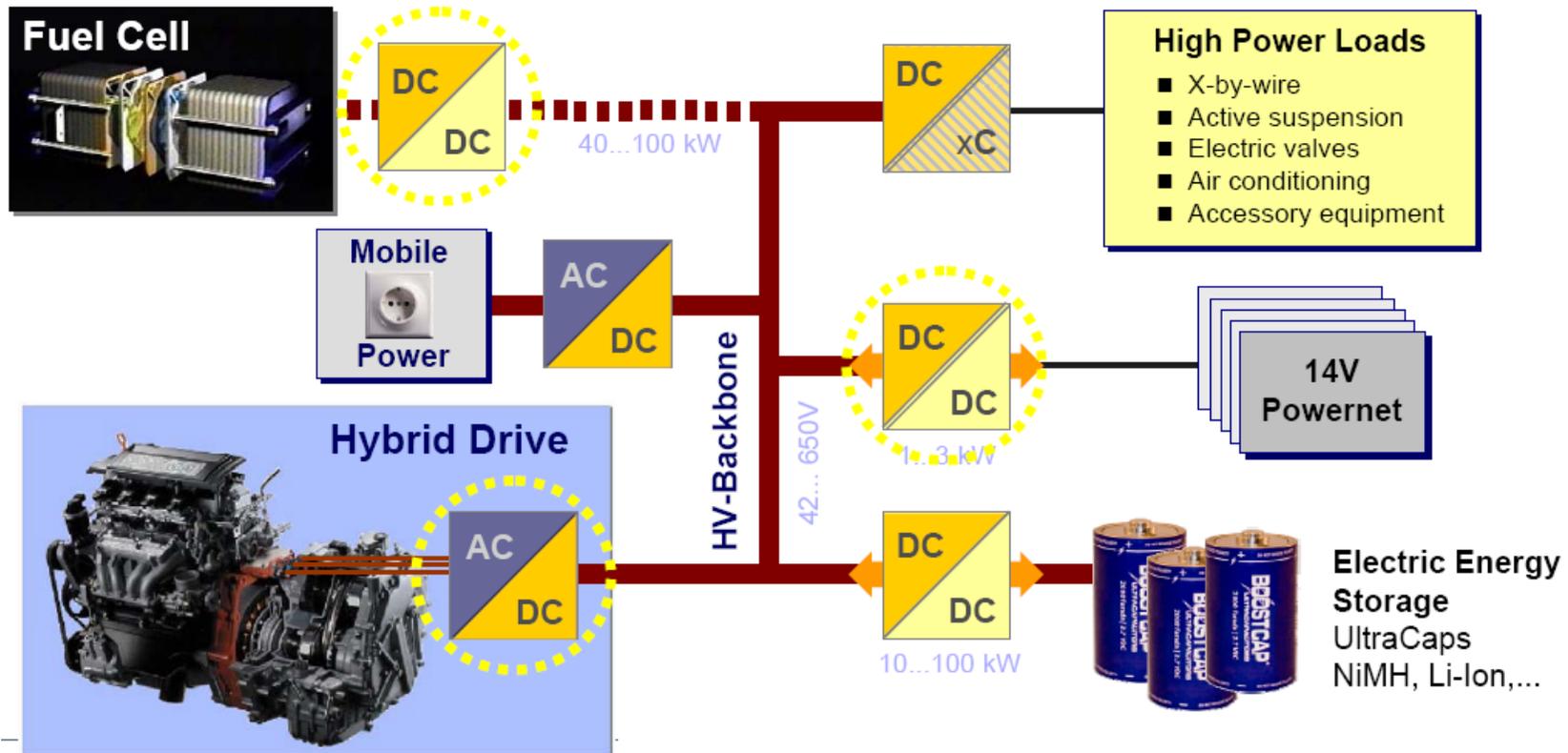
supported by

Eckhard Wolfgang
and *Roadmap Team Automotive*

Outline

- ▶ **General Considerations**
- ▶ **Si / SiC Inverter**
- ▶ **Non-Isolated DC/DC Converter**
- ▶ **Isolated DC/DC Converter**
- ▶ **High Temperature Gate Drive**
- ▶ **Optimization**

Power Electronic Key Systems for the Cars of Tomorrow



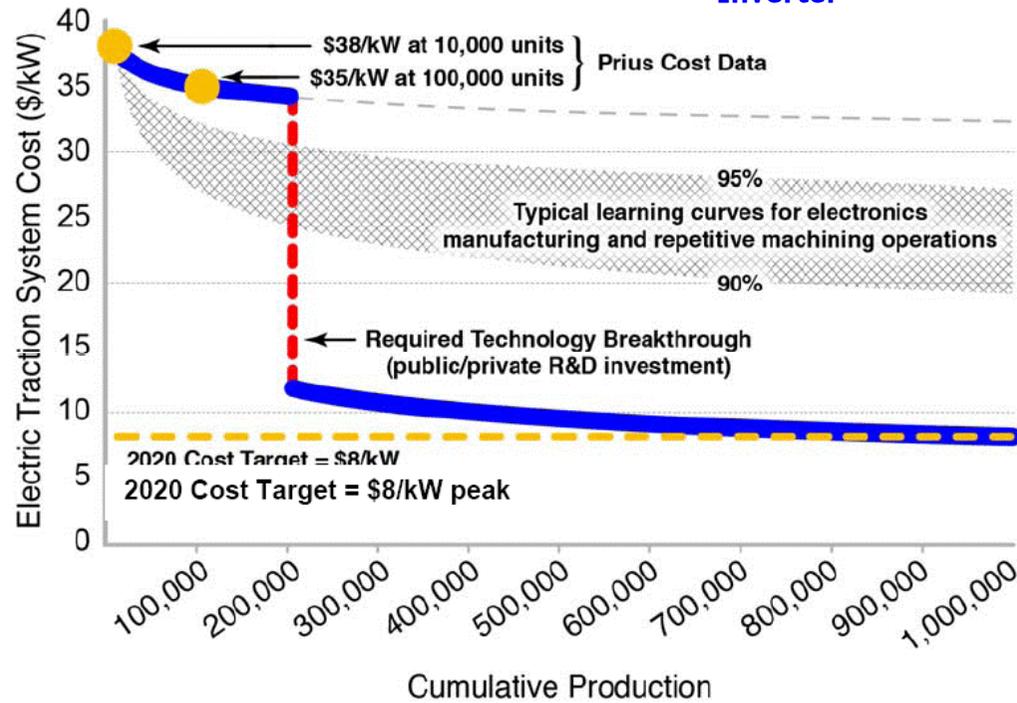
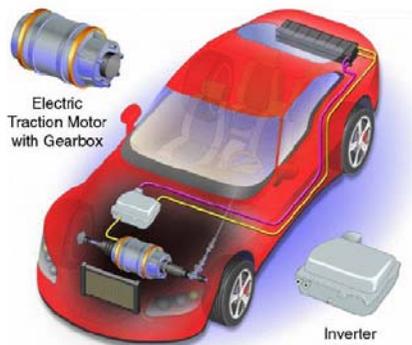
More Electric Car



Market Challenges

2015 Cost Target $\$12/\text{kW} \times 55\text{kW} = \660

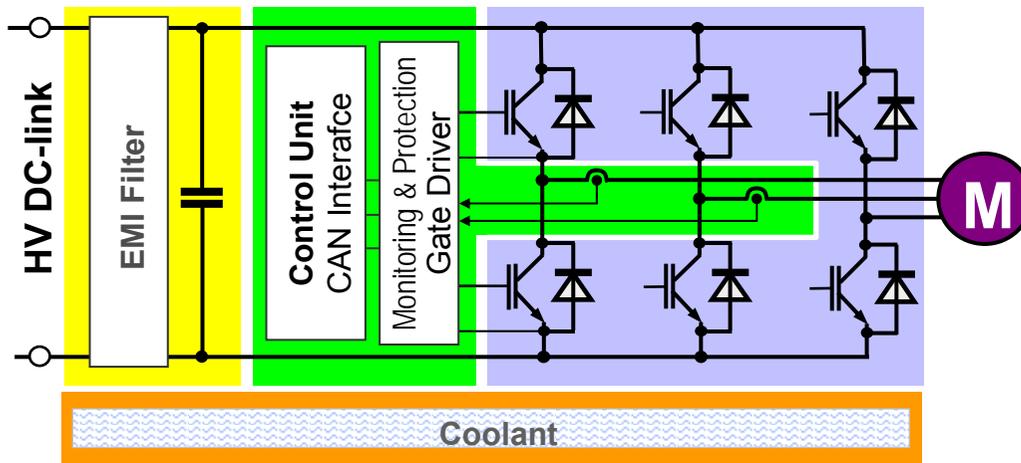
**Electric Traction Motor
Gearbox
Inverter**



Inverter

———
**Topologies
DOF for Optimization
Technologies**
———

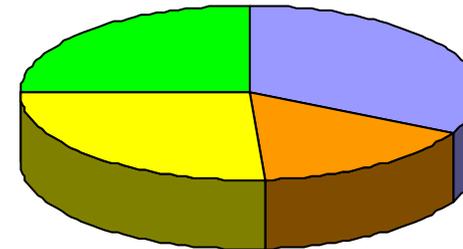
Electric Drive for Hybrid Traction



Typ. Inverter Cost Split

Control, Sensors, and Gate Driver

Power Semiconductors



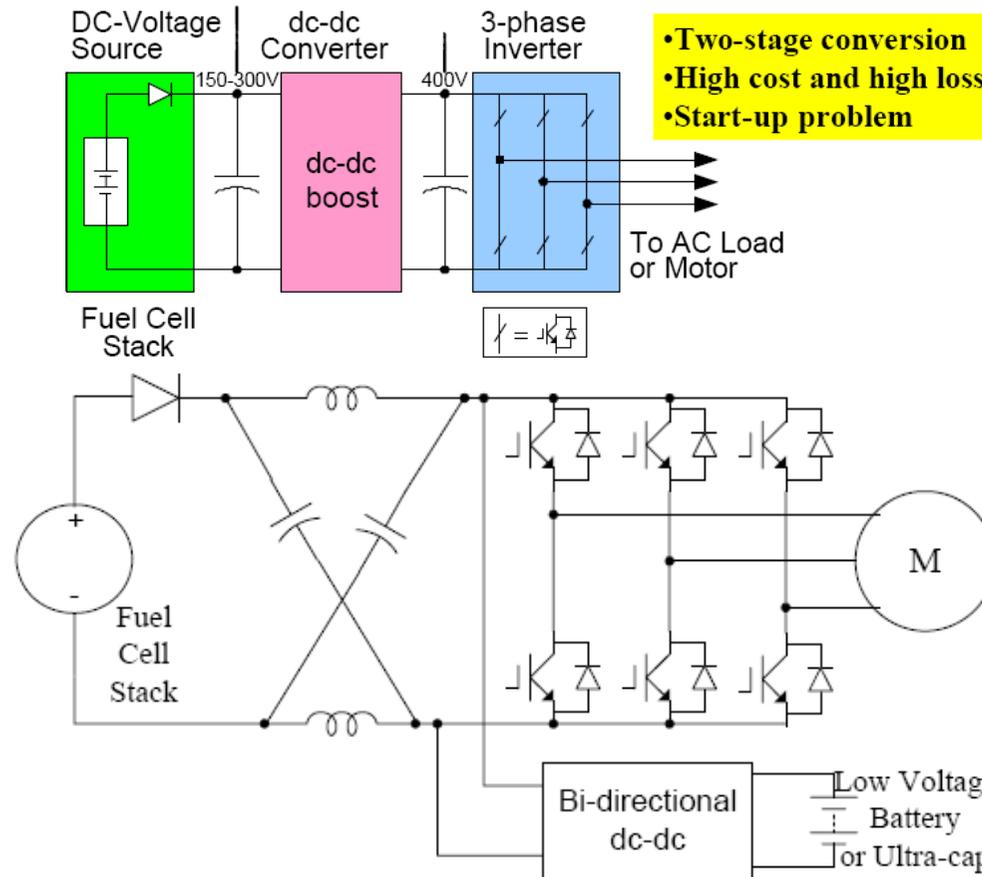
DC Link and EMI Filter

Mounting and Cooling of Power Semiconductors

Alternative Topologies

- ▶ Z-Source Inverter
- ▶ Current DC Link Inverter
- ▶ Matrix Converter

Z-Source Inverter



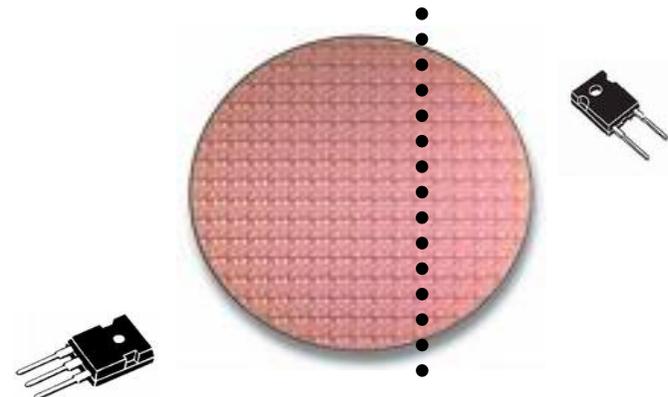
DOF for Optimization

- ▶ Adapted Doping Profile
- ▶ **Partitioning of Total Si Area**
- ▶ **DC Voltage Level ($P = U \cdot I$)**
- ▶ Modulation Concept
- ▶ Output Frequency ($P = M \cdot \Omega$)
- ▶ Switching Frequency

- ▶ **Semiconductor Technology**
- ▶ Coolant Temperature
- ▶ **Cooling Concept**
- ▶ **Temperature Swing**
(Cycles to Failure)

- ▶ **Gate Drive**

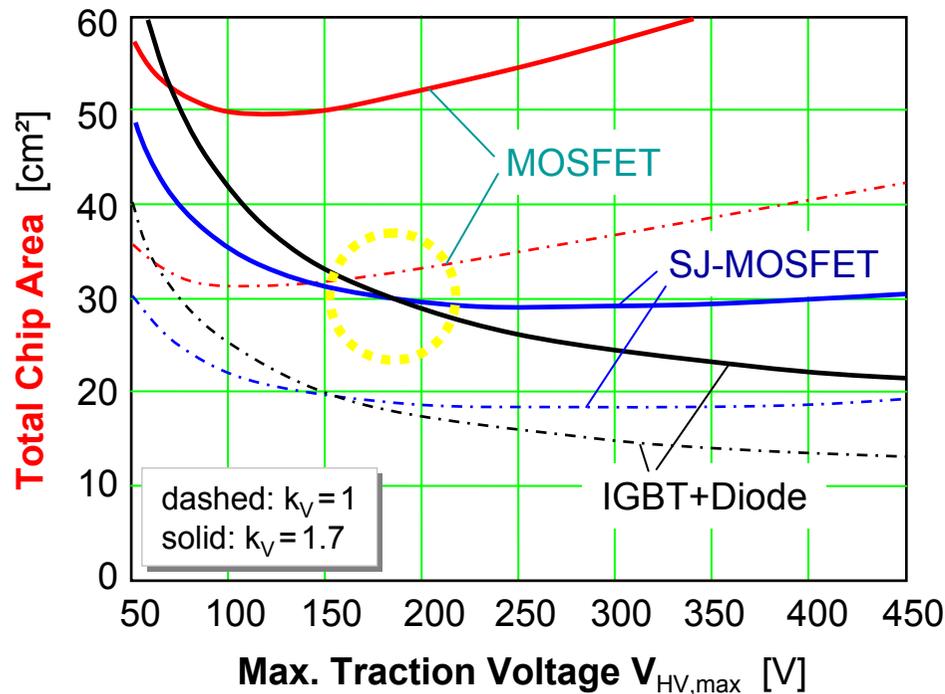
- ▶ **Packaging / Integration**
(ECPE Demonstrator)



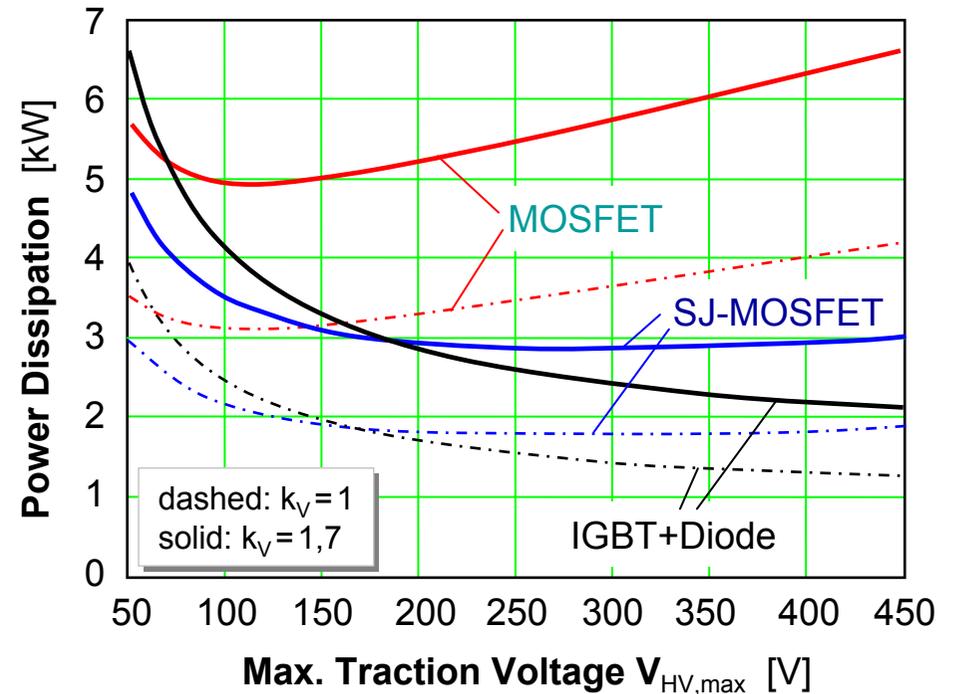
Optimization on System Level

Traction Drive Inverter

Total Power Semiconductor Needs

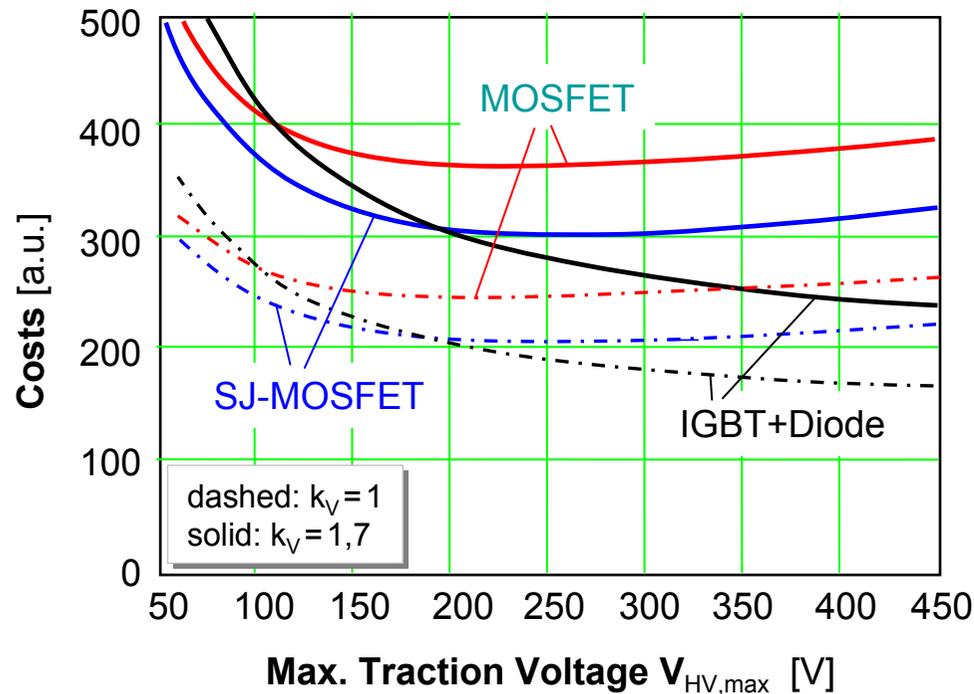


Total Inverter Losses



Traction Drive Inverter

Total Material Costs

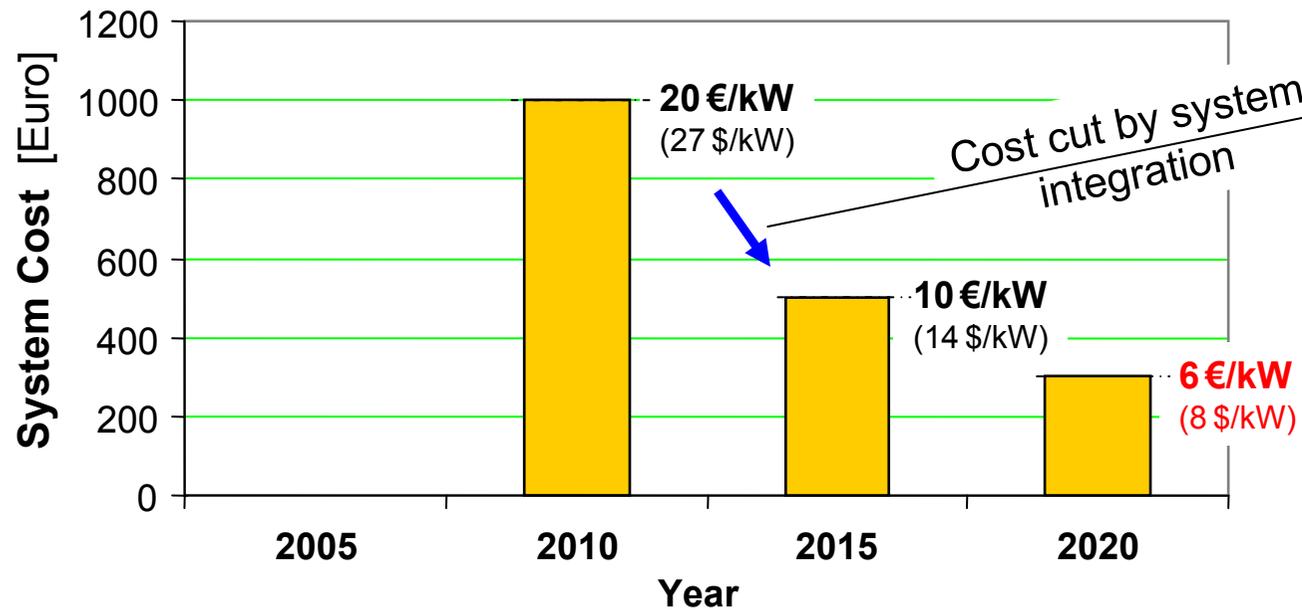


Results

- IGBT is the preferred technology for traction voltages above about 150V
- Total inverter cost, package volume, and losses decrease with increasing traction voltage when using IGBTs
- The inverter becomes considerably less expensive in the case of a constant traction voltage ($k_v=1$)

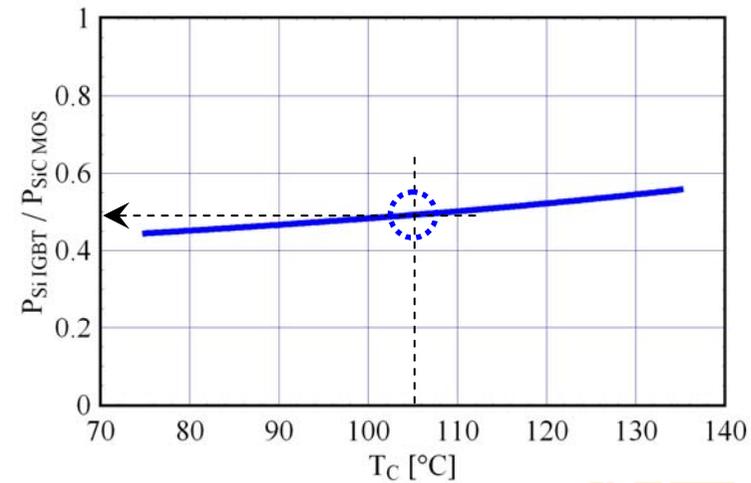
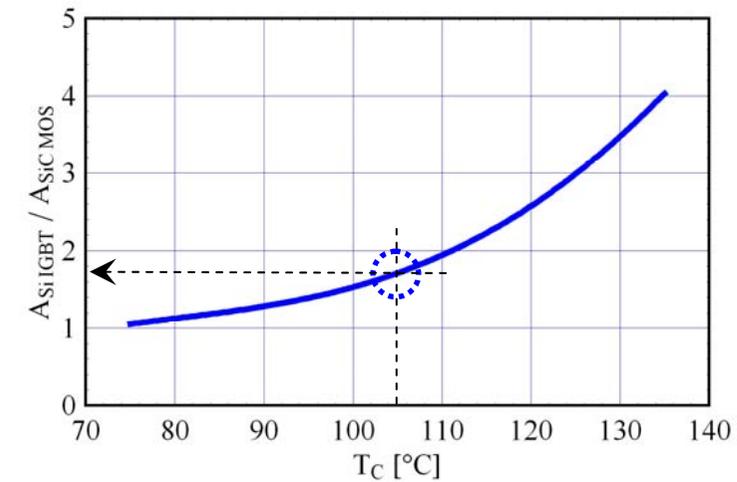
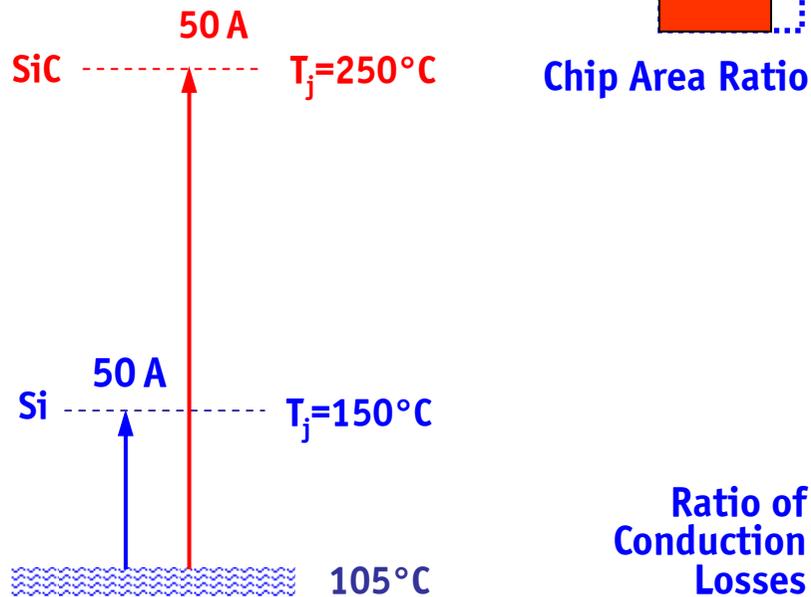
Electric Drive for Hybrid Traction

System Cost Targets



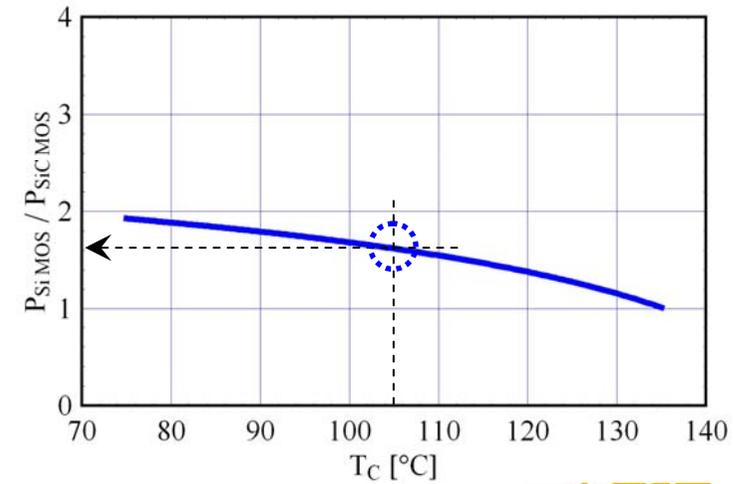
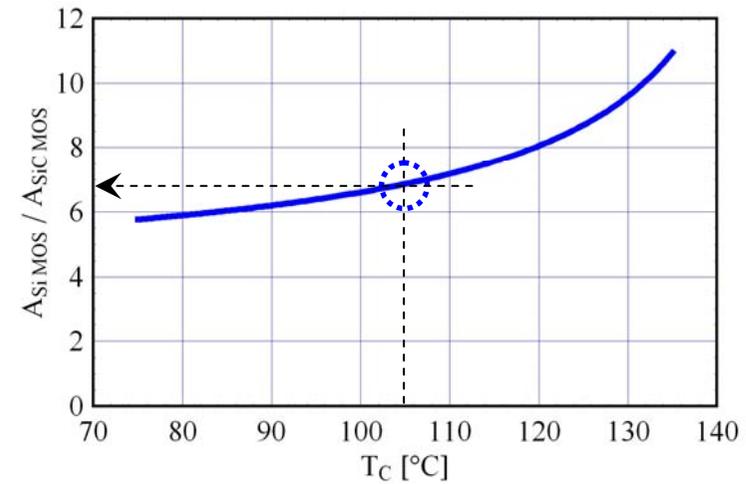
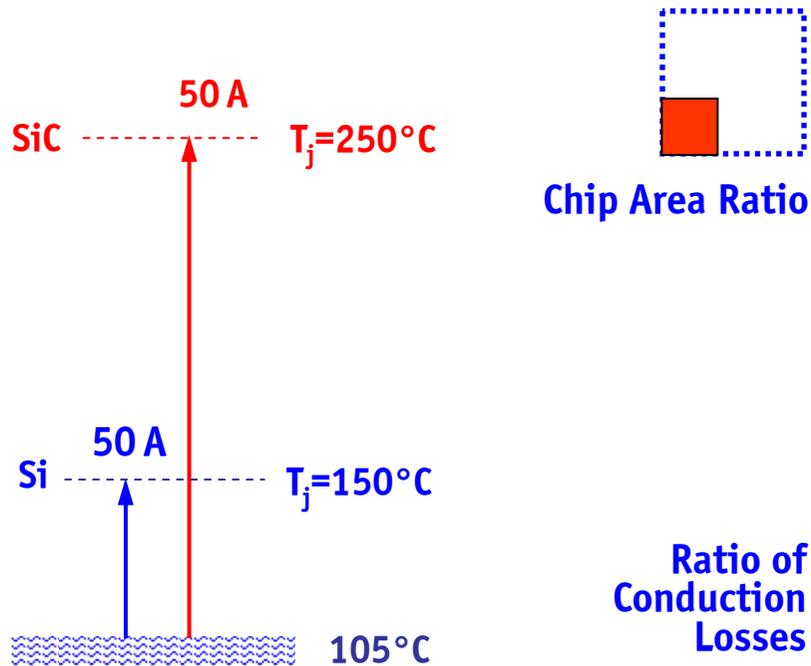
Comparative Evaluation of SiC for 6-Switch Motor Inverters

Trench IGBT 1200V-50A
SiC MOSFET 1200V-50A (CREE)



Comparative Evaluation of SiC for DC/DC Converter

Si CoolMOS C3 1200V-50A (Extrapolated)
 SiC MOSFET 1200V-50A (CREE)

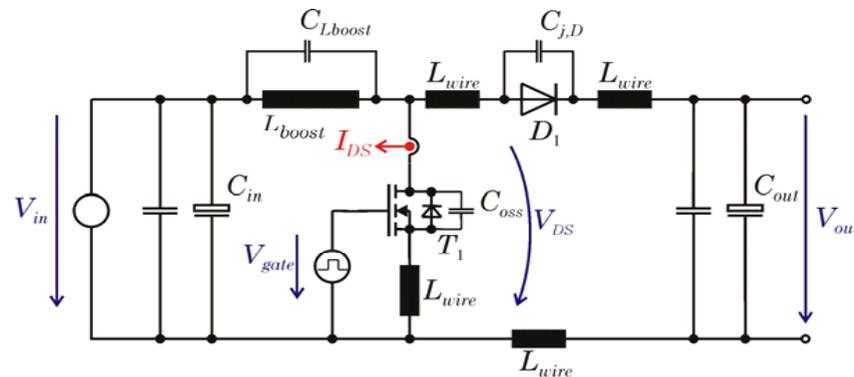


Switching Transient Shaping

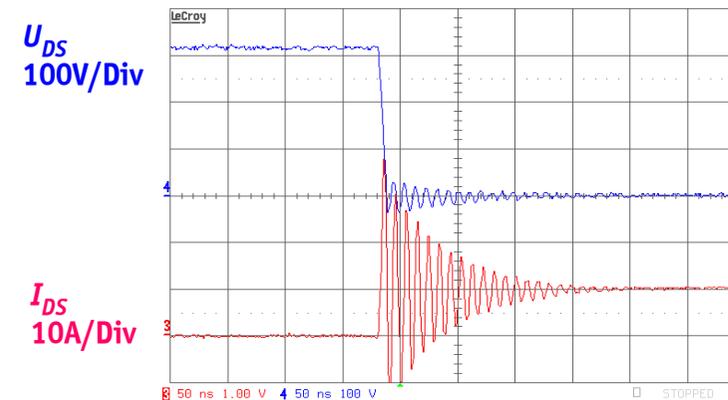
- ▶ Minimization of Parasitics



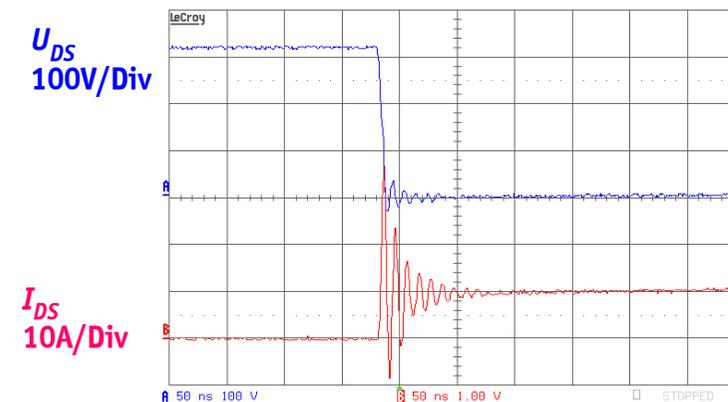
- ▶ Passive Damping
- ▶ Gate Drive / Active Damping



Without Damping Layer



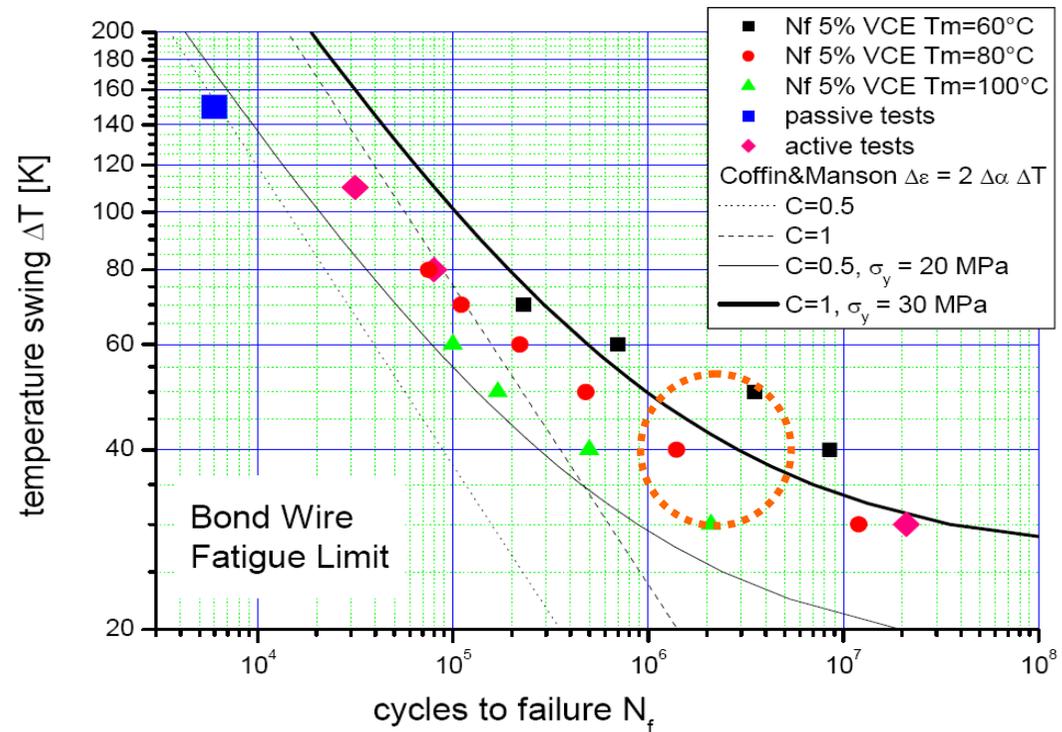
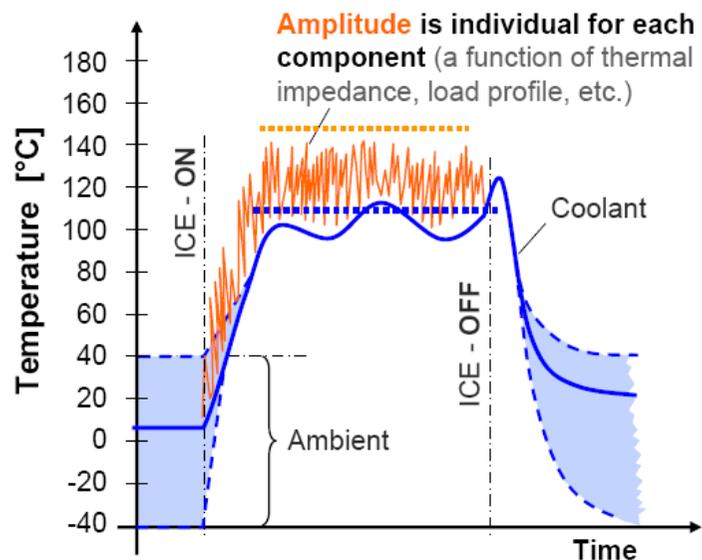
50ns/Div PCB Damping Layer



Thermo-Mechanical Reliability

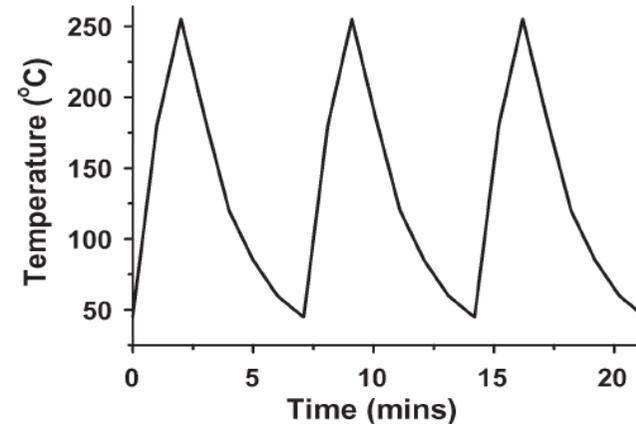
- ▶ **Passive Cycles** 15'000
- ▶ **Active Cycles** > 3'000'000

Bond Wire Fatigue Limits

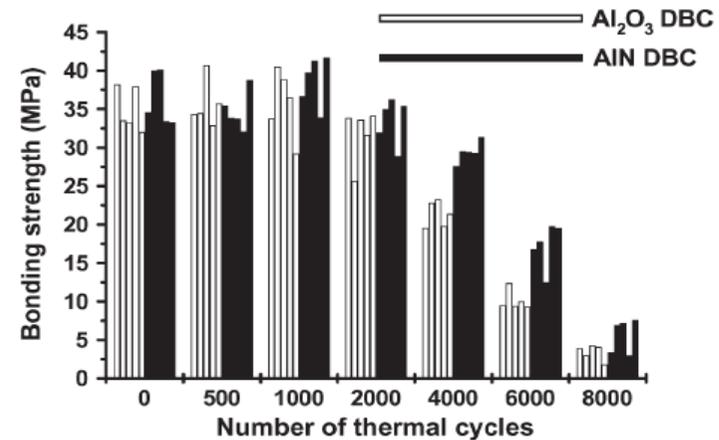
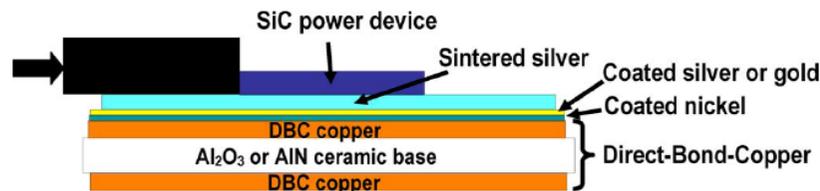


Thermo-Mechanical Reliability

- ▶ SiC Power Device Assembly
- ▶ Low Temperature Sintered Silver Die Attachment
- ▶ Thermal Cycling 50°C 250°C
- ▶ 6'000 TC Survived



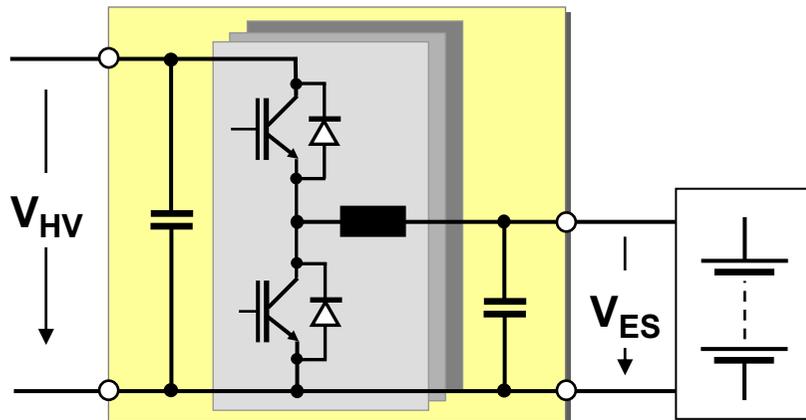
Die-Shear Test



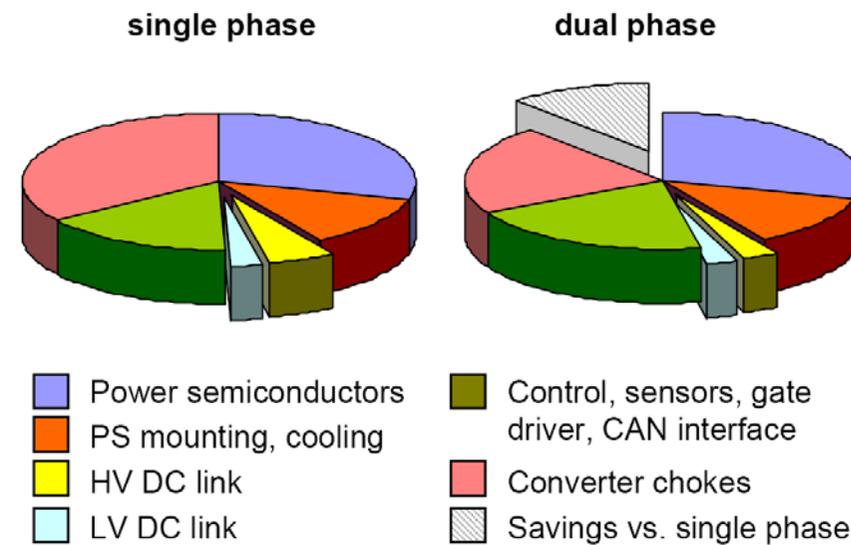
Non-Isolated DC/DC Converter

—— Overlapping Input/Output
Voltage Ranges ——

Traction Voltage Converter

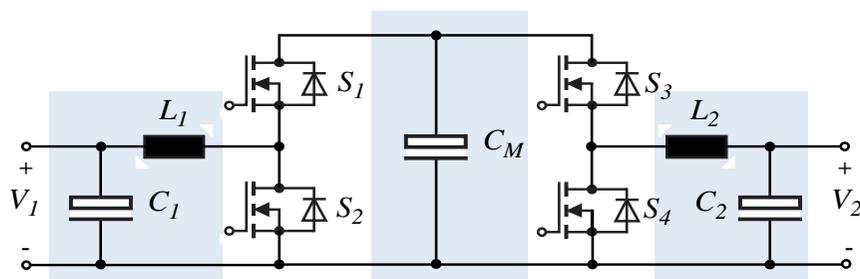


Typ. Converter Cost Split



Bi-Directional DC/DC Converters for Overlapping Voltage Ranges

Cascaded Boost-Buck Converter

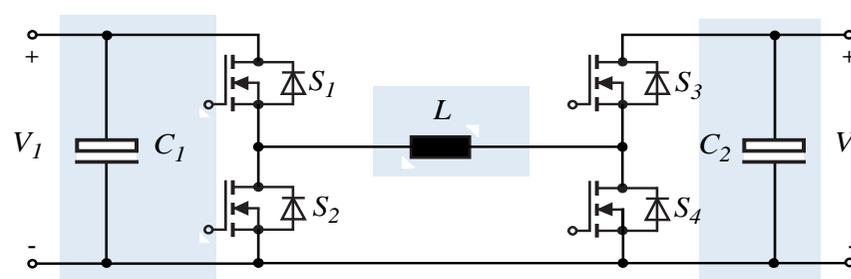


Large Passive Components Count

- 3 Capacitors
- 2 Inductors



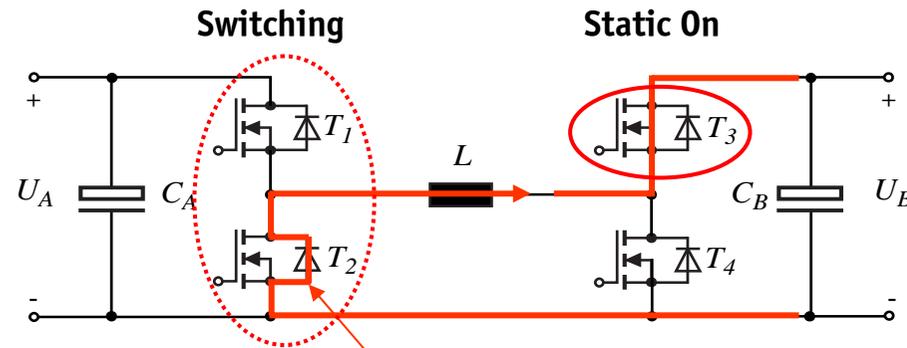
Cascaded Buck-Boost Converter



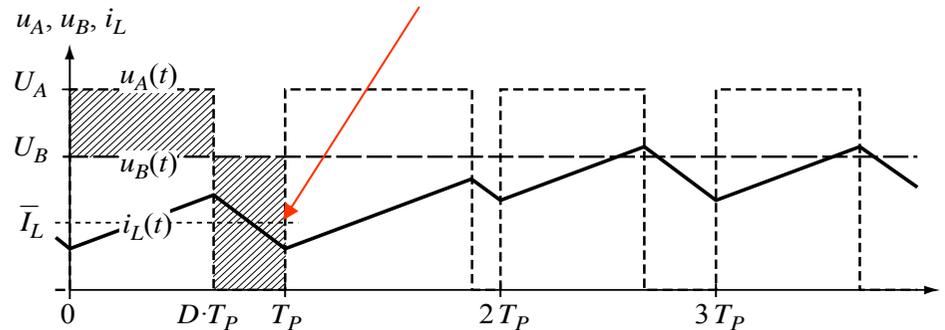
Minimum Passive Components Count

- 2 Capacitors
- 1 Inductor

Cascaded Buck-Boost Converter



Diode reverse recovery losses



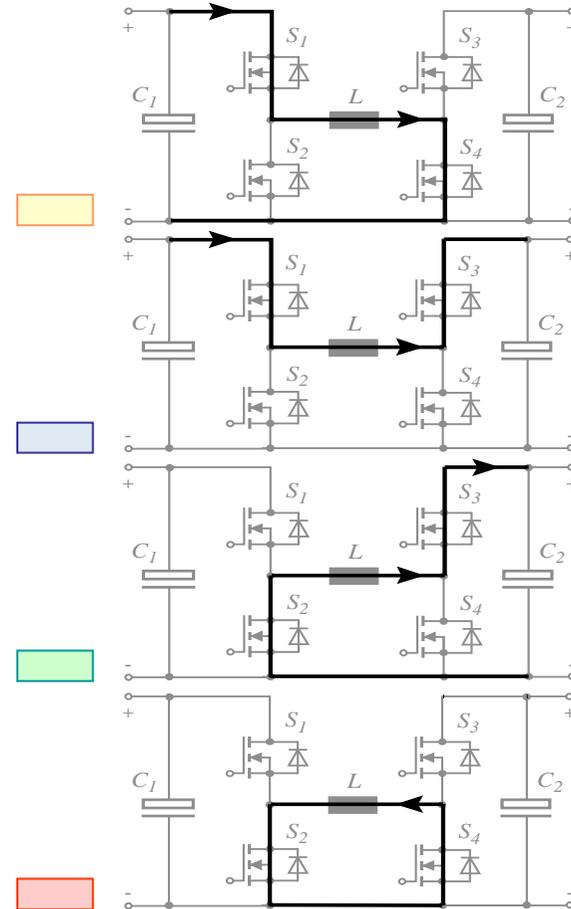
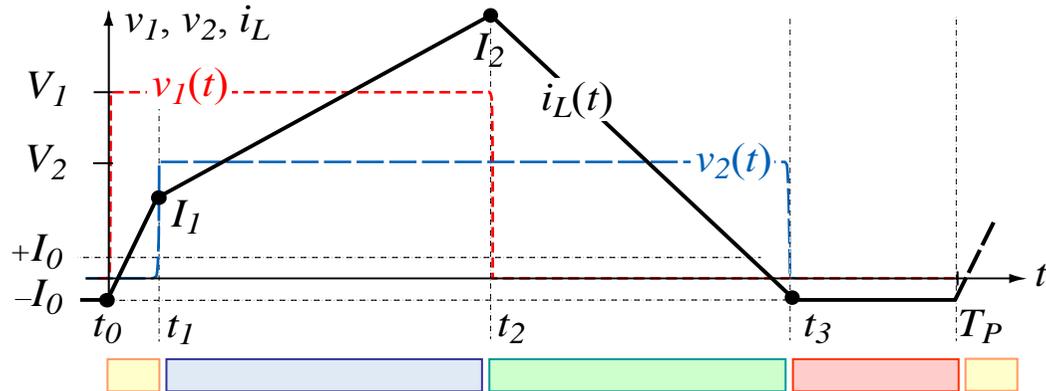
Methods to Reduce Switching Losses

- ▶ Silicon Carbide (SiC)
- ▶ Soft-Switching - ZVS, ZCS

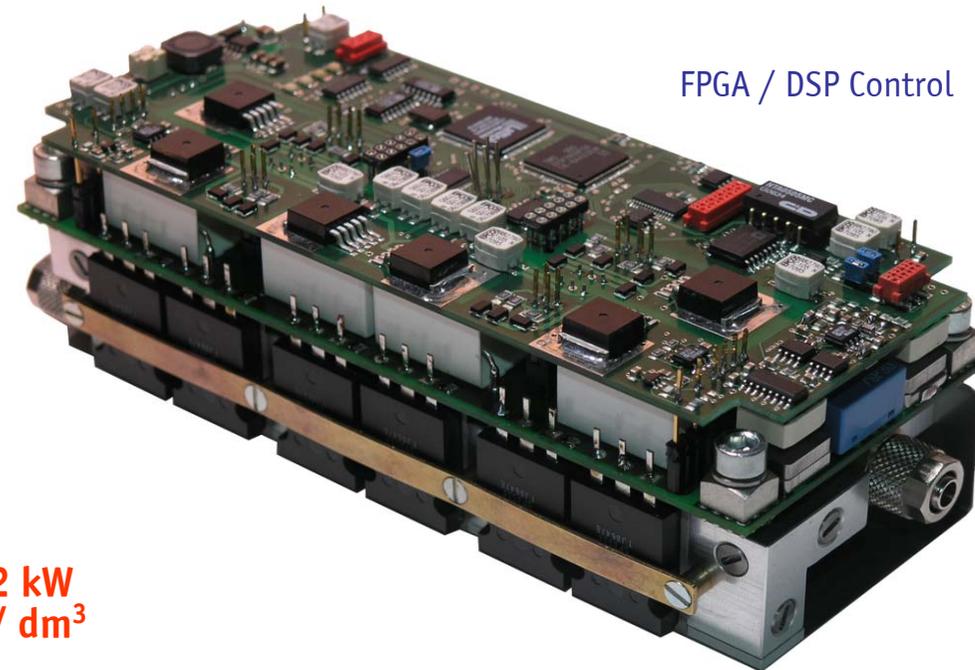
Low-Loss Modulation

Operating Modes

Buck operation: $V_2 < V_1$, side 1 \rightarrow side 2



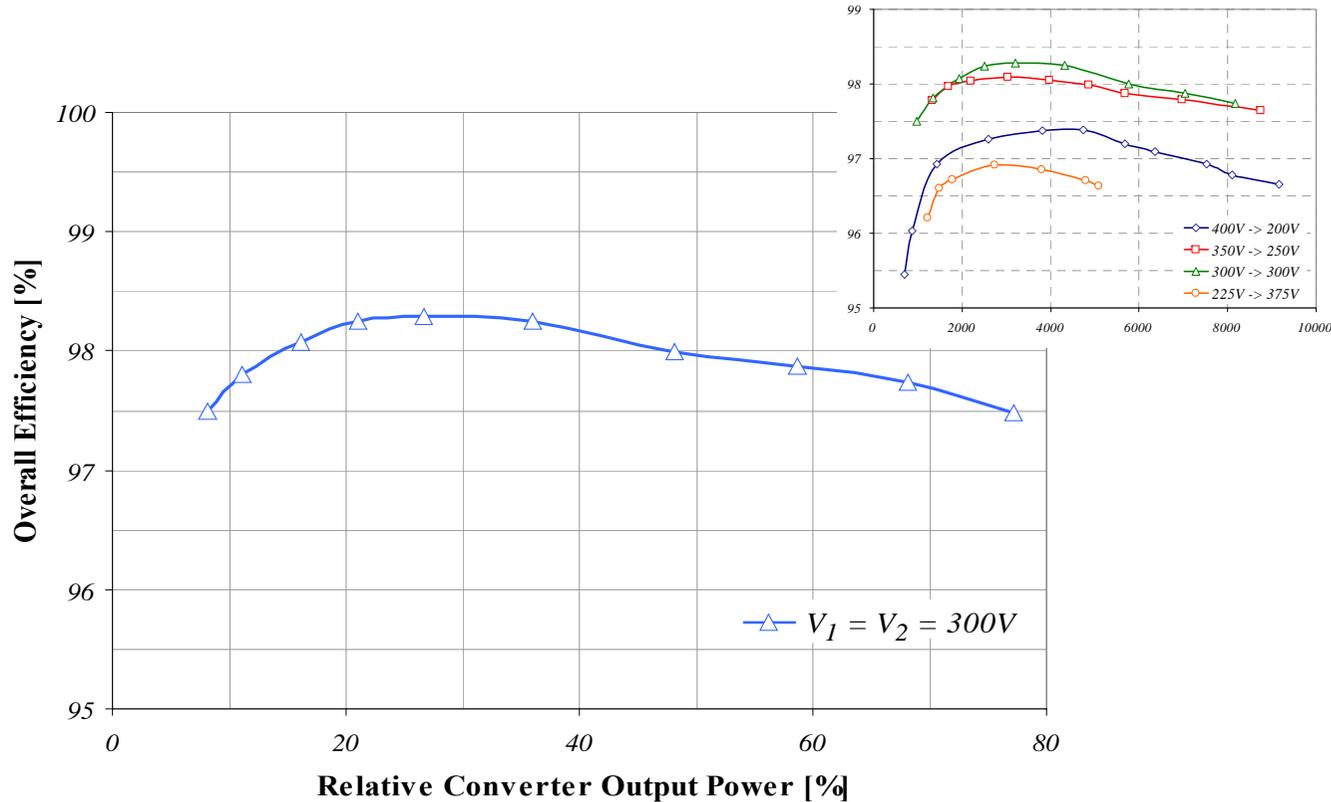
Converter Module Hardware



FPGA / DSP Control

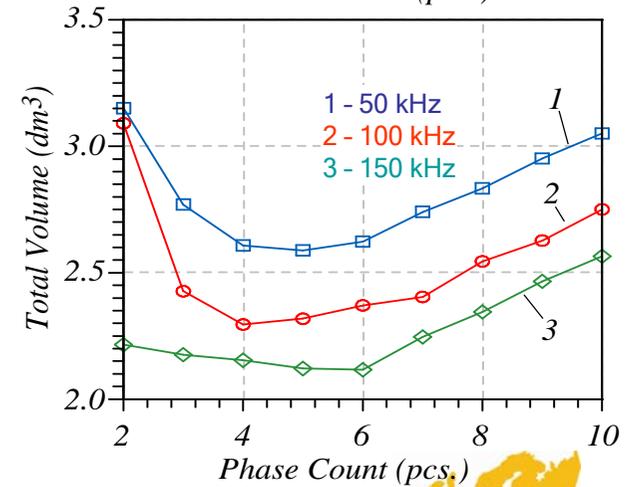
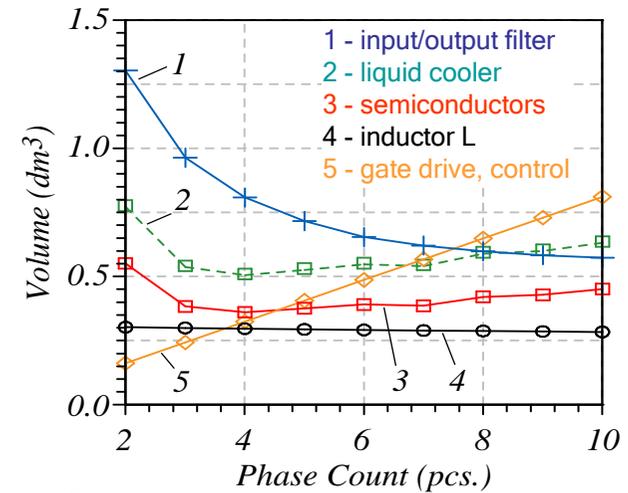
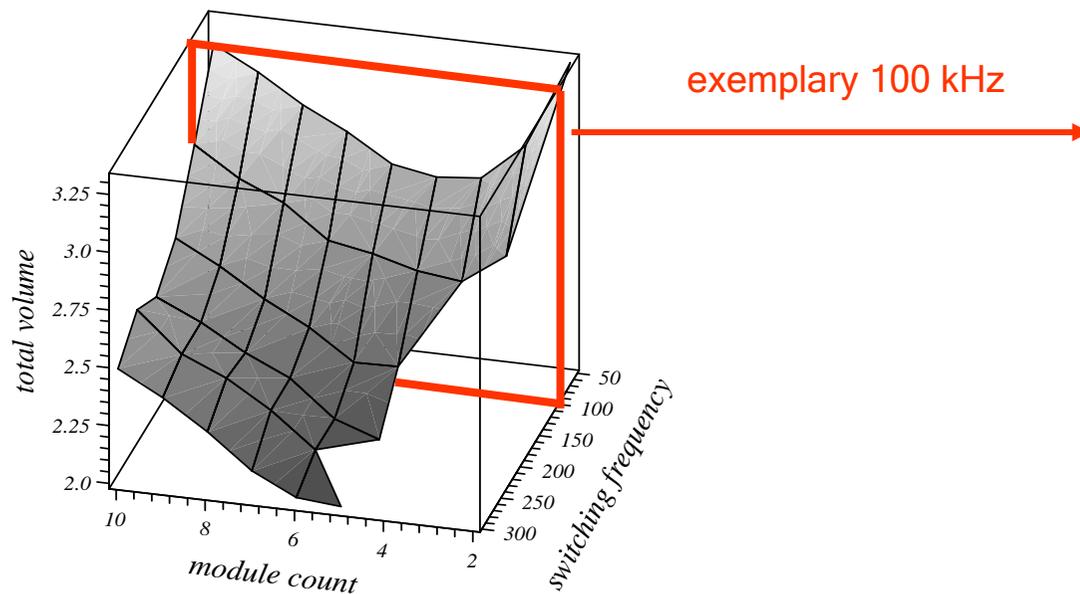
- ▶ Peak Power Rating 12 kW
- ▶ Power Density 17.5 kW / dm³

Overall Efficiency vs. Output Power

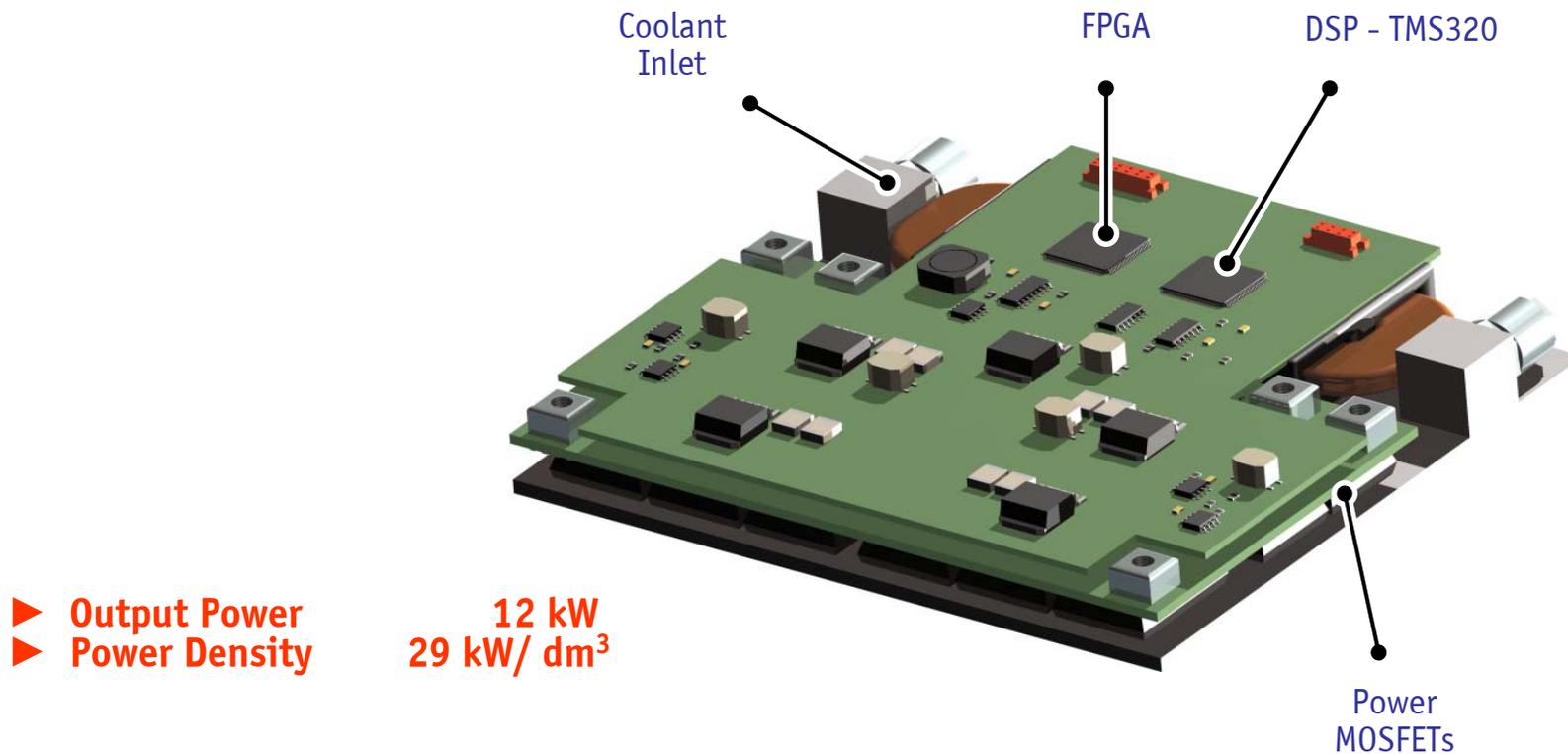


Converter Volume Optimization

- ▶ **Module Count** 2 .. 10
- ▶ **Switching Frequency** 50 .. 300 kHz



Ultra-Compact Converter Module



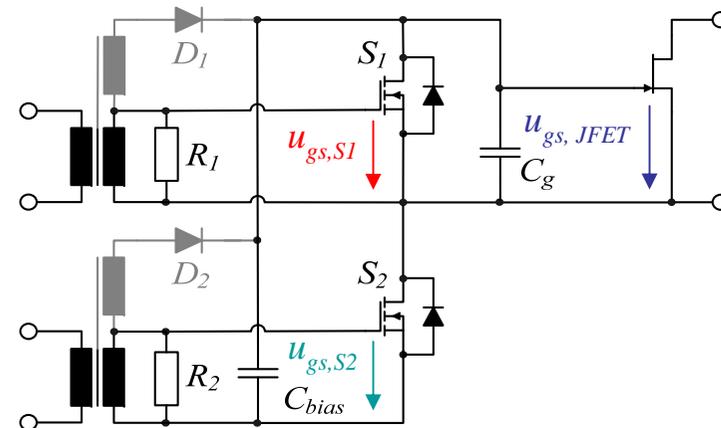
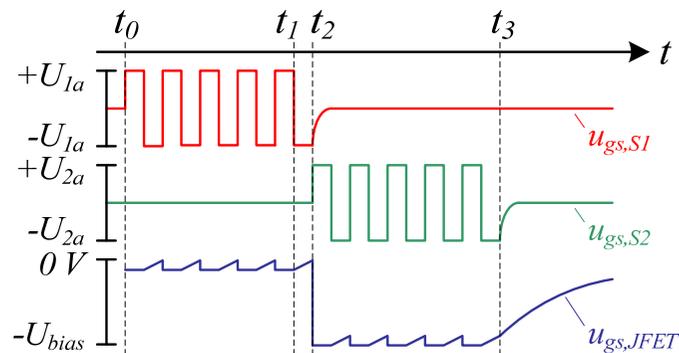
Isolated High Temperature SiC J-FET Gate Drive Circuit

$T_a = 250^\circ\text{C}$



Phase Difference Circuit

Proposed by D.C. Hopkins, Univ. at Buffalo, USA

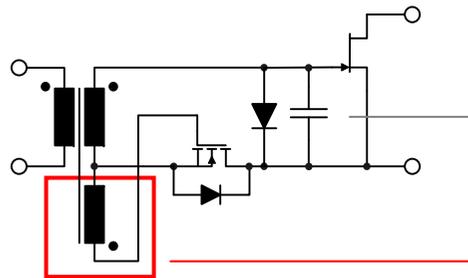


- ▶ Vs Product: Bipolar transformer output voltage
- ▶ Capacitor C_g to preserve JFET gate voltage during MOSFET S_1 or S_2 Off-Time

Advantages and Drawbacks

- No Duty-Cycle limitation (static Turn-Off)
- High switching speeds (MOSFET half-bridge)
- High complexity
- High costs

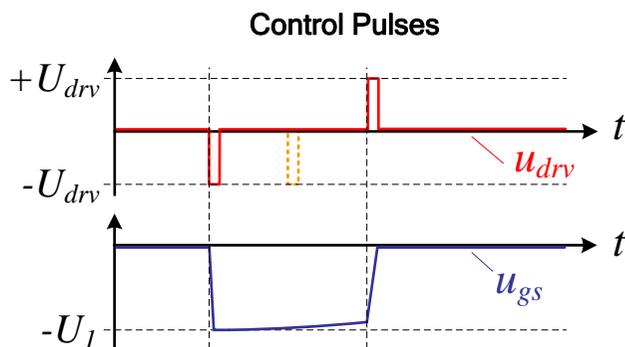
Edge-Triggered Driving Circuits



Size of Capacitor C_g

- Large capacitances reduce switching speed
- Large capacitances cause significant losses
- Small capacitances limit Off-Time

Second winding due to auxiliary switch U_{gs} limits

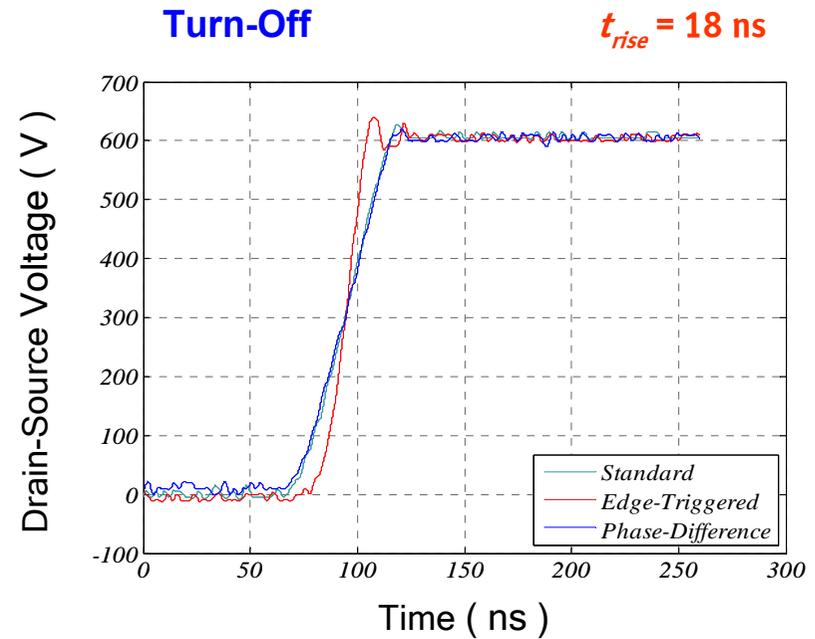
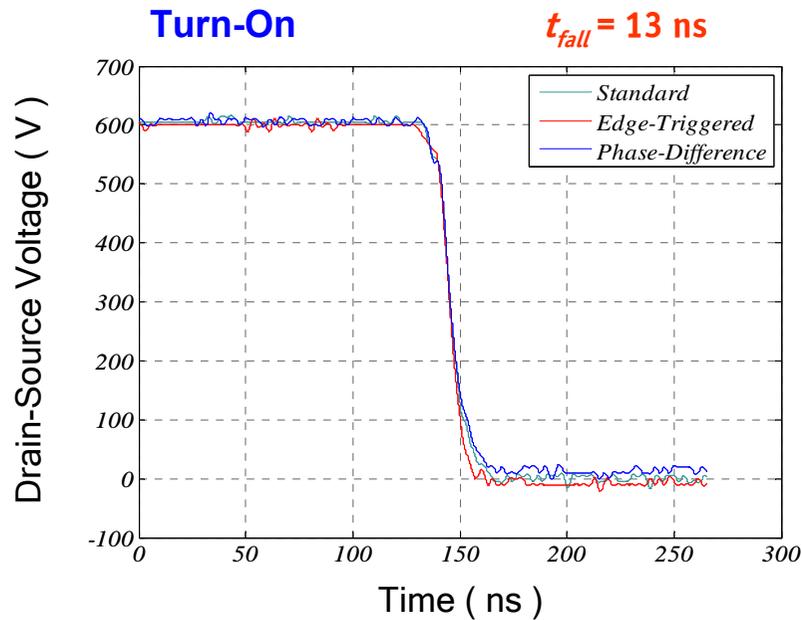


Advantages and Drawbacks

- Moderate Active Component Count
 - High Switching Speeds
 - Large Duty-Cycle Range (1% ... 100%)
 - (Off-Time limited by capacitor size)
- special pulse pattern to provide negative bias useable

Experimental Results

Performance Comparison

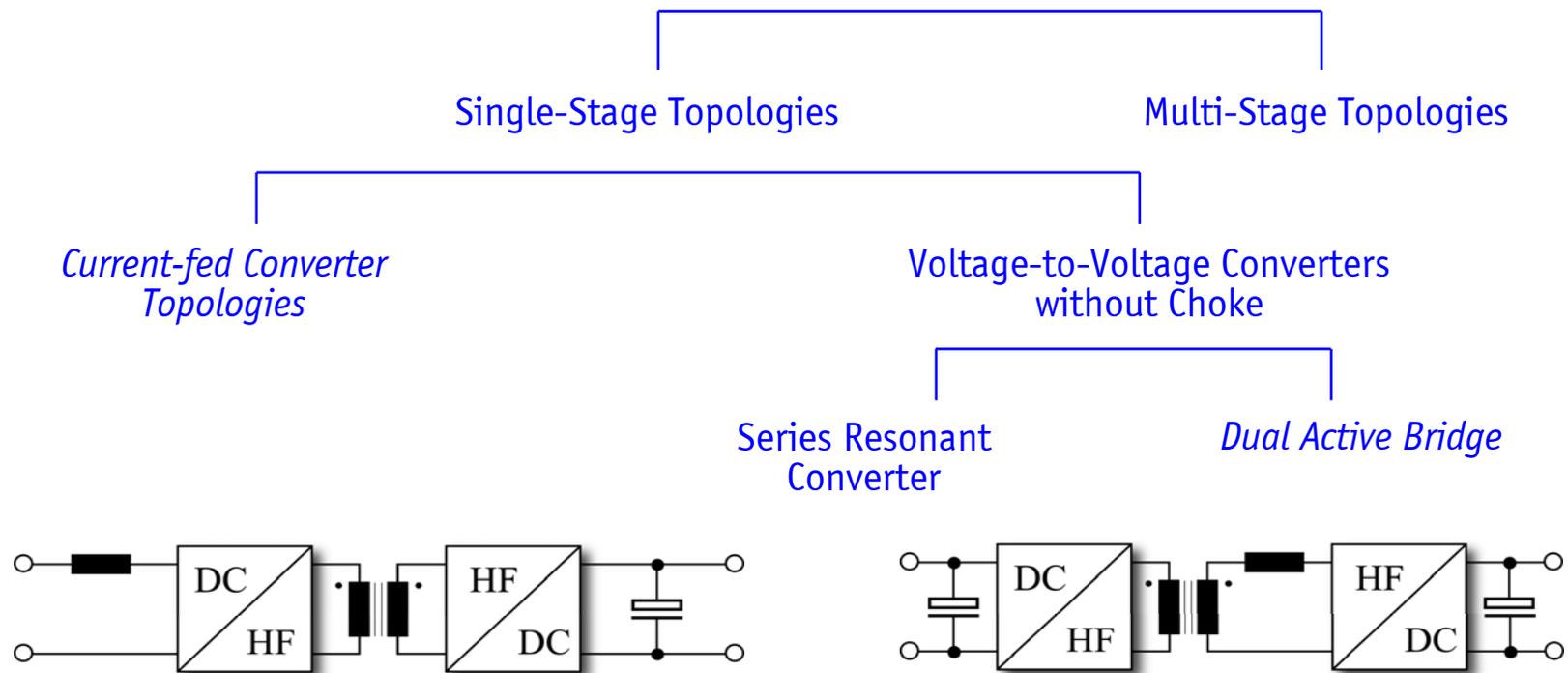


► Edge-Triggered Circuit shows Excellent Performance

Isolated DC/DC Converter

—————
**Dual Active Bridge
Magnetically Integrated
Current Doubler**
—————

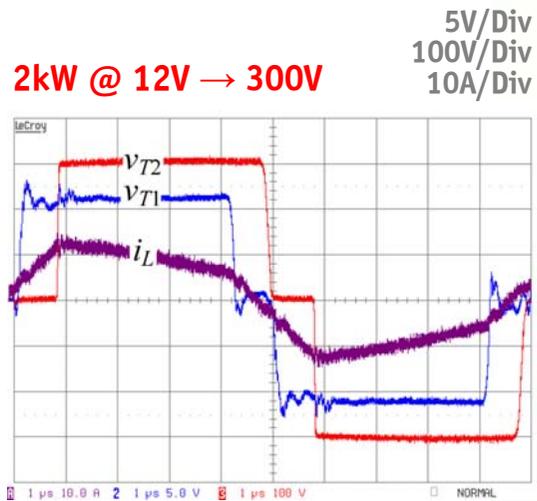
Isolated Bi-Directional DC/DC Converter Topologies



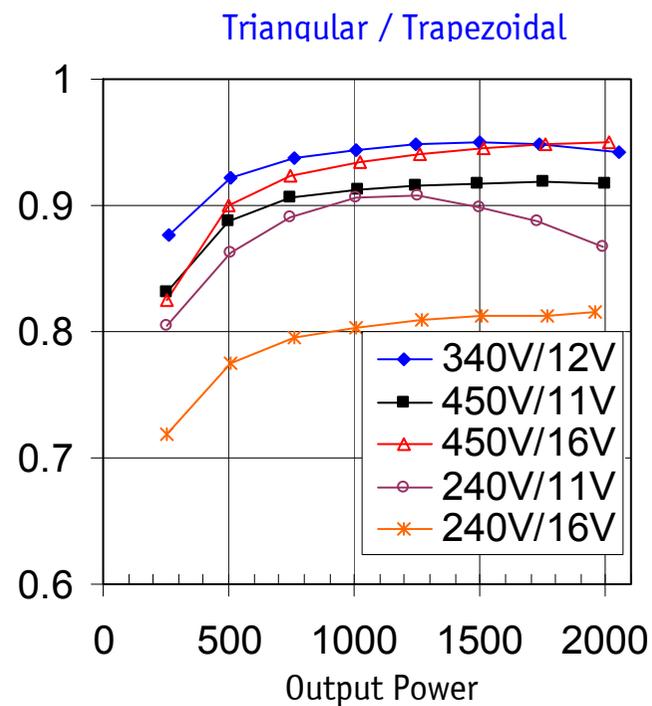
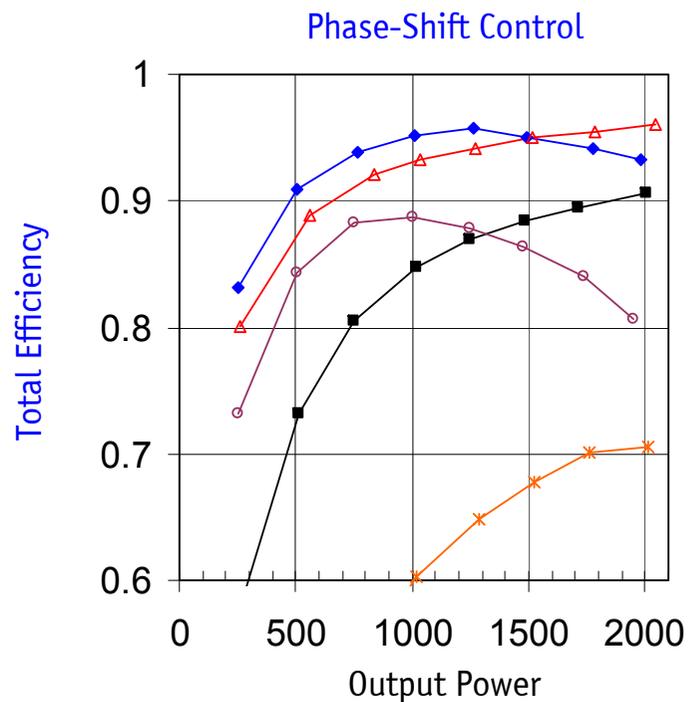
► No High Current Inductor

Prototype of the Dual Active Bridge

2kW
11...16V → 220...450V
 $\eta > 90\%$
100kHz
2 kW/dm³



Experimental Results

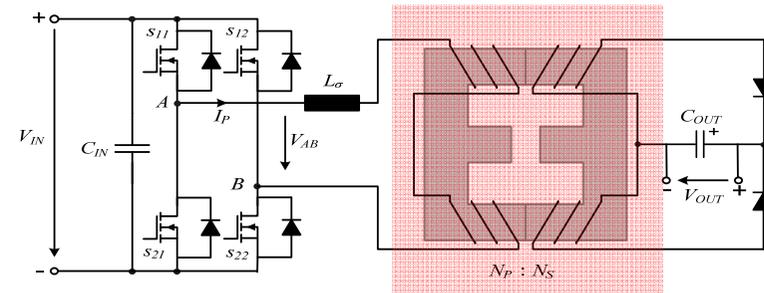
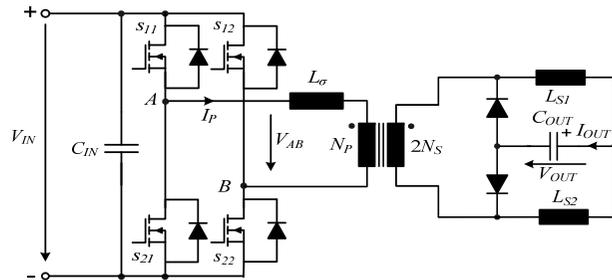


- ▶ Efficiency Increased by 10% at 2kW Output
- ▶ Significantly Higher Efficiency at Partial Load

Isolated DC/DC Converter

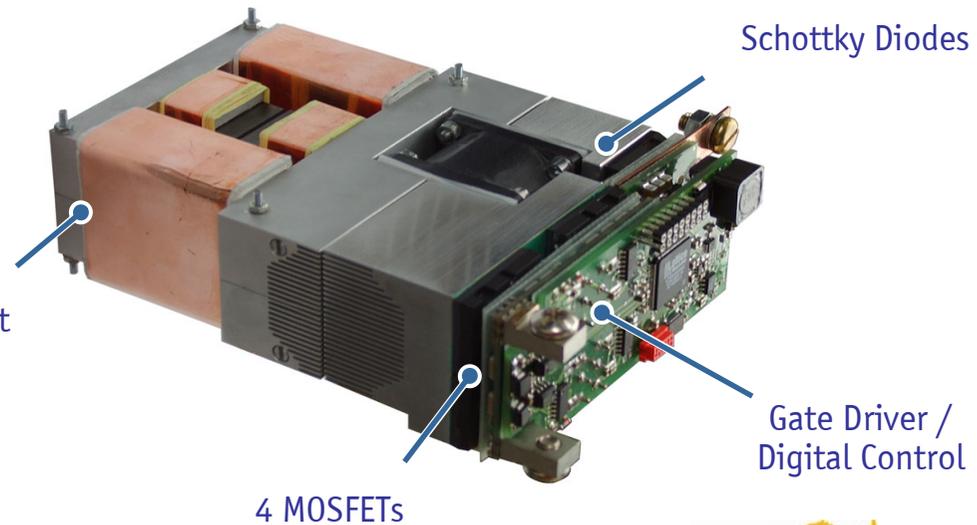
_____ Magnetically Integrated
Current Doubler _____

Current Doubler with Integrated Magnetics



Output Power 5kW
Switching Frequency 200kHz

► **Power Density 8.7 kW/dm³**



Enabling Technologies Identified in Copenhagen Roadmap Meeting

- **Advanced Cooling of Power Semiconductors**
 - **Increased Thermal Cycling Capability / Increased ΔT_{j-c}**
 - **Advanced Packaging Materials**
 - **Advanced Cooling of Passives**
 - **High Current Low HF Loss Interconnection Technologies**
 - **Local EMI Shielding / Filtering**
 - **Integration of Gate Drives and Sensors etc.**
 - **Reliability / Robustness Test Procedures**
-
- **Multi-Domain Design / Optimization Platform**



System Optimization

—————
Pareto-Optimal Design
Technology Vectors
Sensitivities
—————

Bottom-Up Roadmap Approach for Power Electronic Systems

► How to Identify Future Key Technologies / Required Progress ?

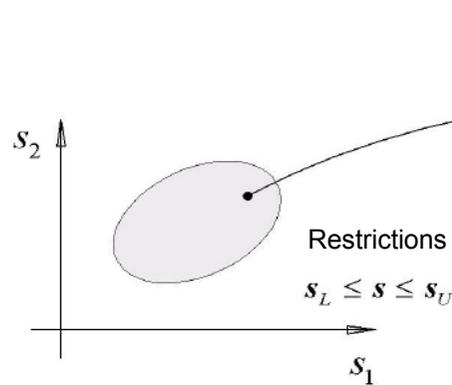
1. Clarify State of the Art & Mapping of Component Technologies into System Performance
Demonstrator Systems
2. Define Goal - as Resulting from Top-Down Analysis
3. Analyze Sensitivities
4. Identify Most Influential Technologies
5. Derive Required Progress in Specific Technology Metrics / FOM

Sensitivities & Technology Vectors

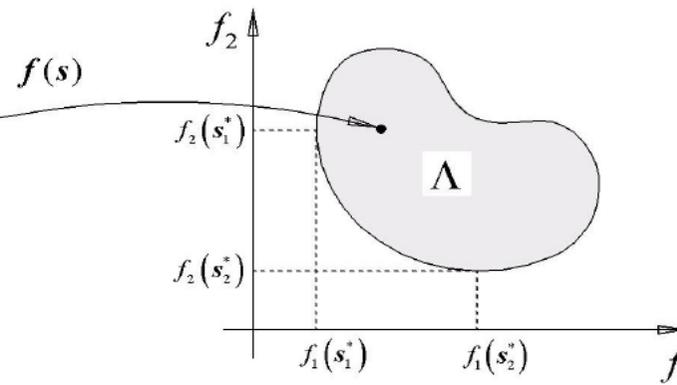
Conflicting Optimization Goals

- ▶ Volume / Weight
- ▶ Efficiency
- ▶ Costs

Technology Space



Performance Space



Pareto-Optimal Solutions in a Convex Region

