

3-Φ AC/AC Matrix & Current Source Converter Modulation & Control

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Part 3 — Tutorial
Emerging Bidirectional
Switches and Their
Impact on Future AC Power
Converter Applications

Oct. 10, 2021



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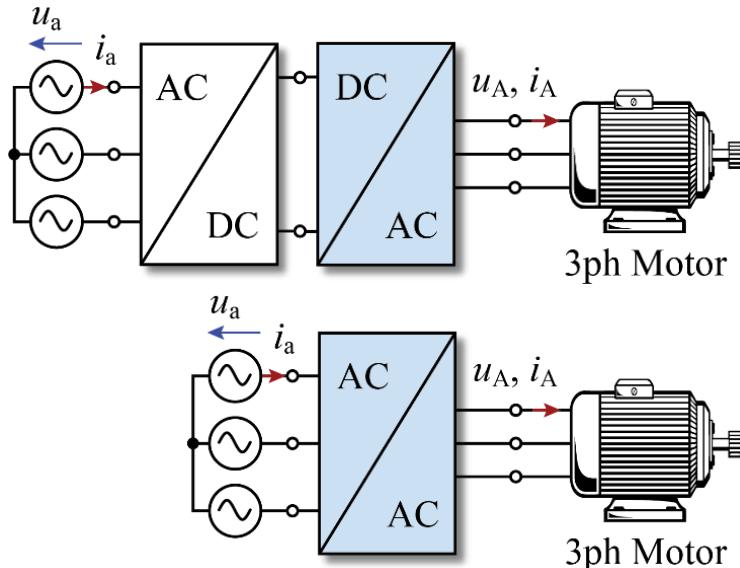
Outline

- ▶ *Introduction*
- ▶ *Indirect & Direct Matrix Converter*
- ▶ *I-DC Link AC/AC Converter*
- ▶ *Comparative Evaluation*
- ▶ *Conclusions*

T. Friedli
F. Krismer
F. Schafmeister
Acknowledgement

Variable Speed Drive Inverter Concepts

- *Voltage OR Current DC-Link AC/AC Converters*
- *Indirect OR Direct AC/AC Matrix Converters*



38%
of electric energy use is for motors
in commercial buildings.

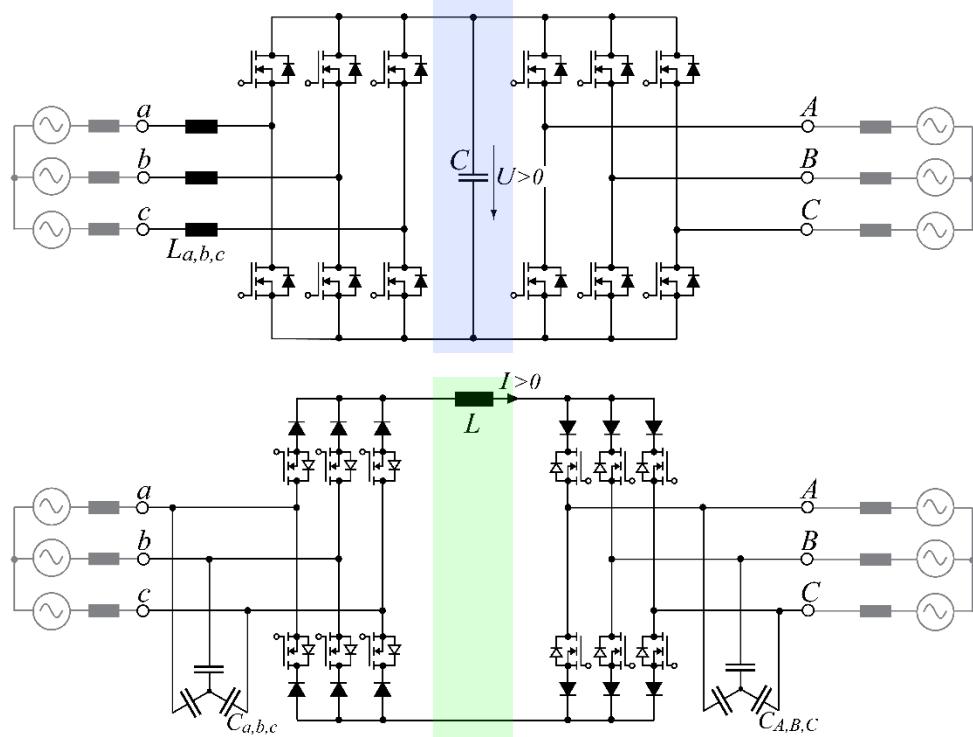
70%
of electricity consumed by industry
is used in electric motor systems.

Source: **ABB**

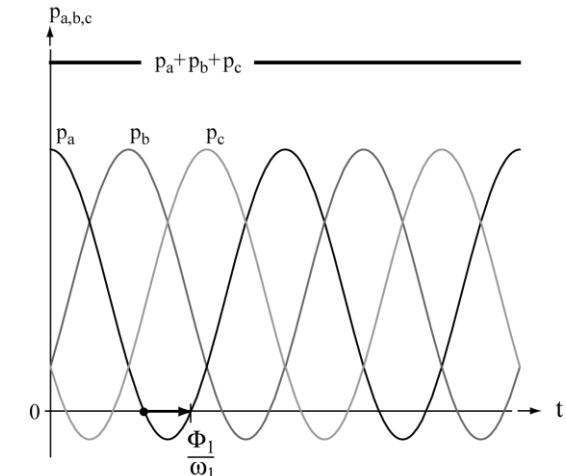
- *45% of World's Electricity Used to Power Motors in Buildings & Industrial Applications*

DC-Link AC/AC Converter Topologies

- *Front-End PFC Rectifier Stage | Load-Side PWM Inverter Stage*



$$P = \frac{3}{2} \hat{U}_1 \cdot \hat{I}_1 \cos \Phi_1$$



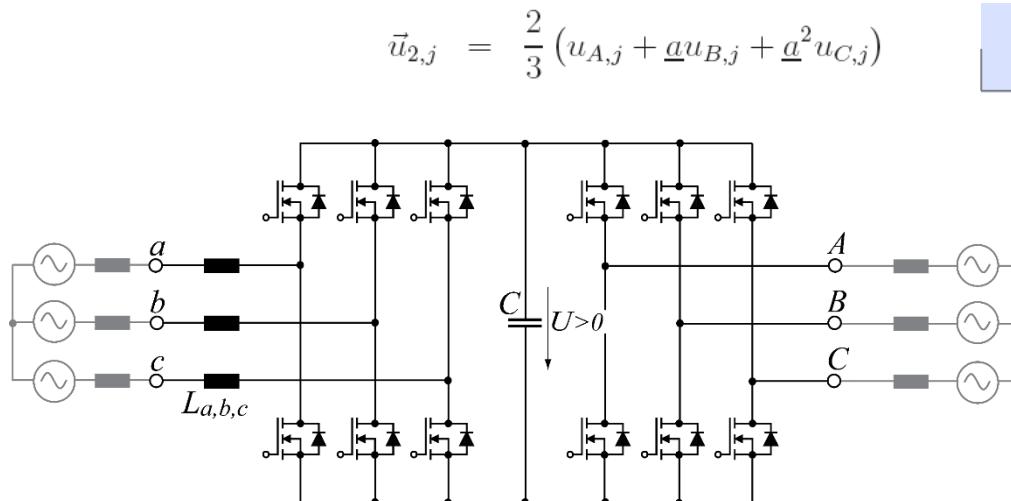
- *Constant 3-Φ Instantaneous Power Flow → No Low-Frequency Power Buffer Requirement*



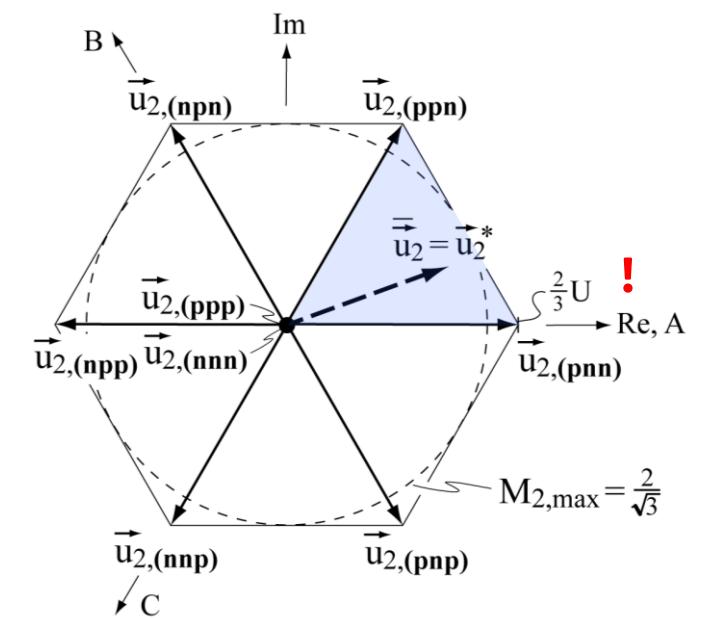
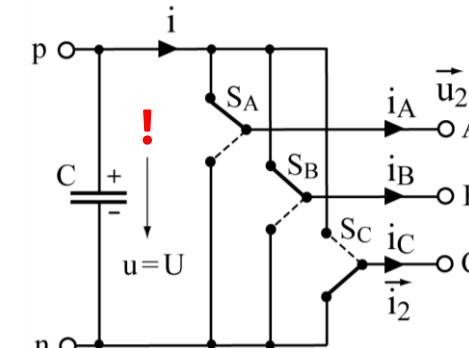
*Voltage Source
Converter*

VSC Space Vector Modulation (1)

- *Switching Considering Interlock Delay Times*
- *$2^3 = 8$ Switching States*



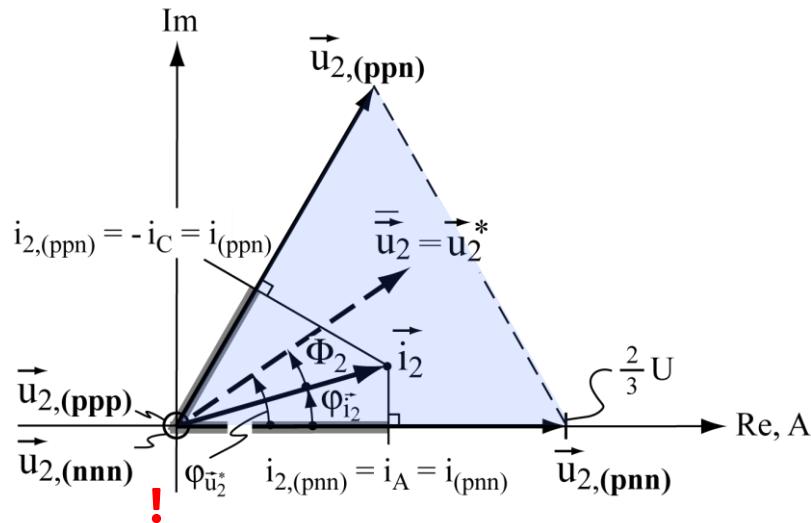
$$\vec{u}_2^* = \hat{U}_2^* e^{j\varphi} = \hat{U}_2^* e^{j\omega_2^* t}$$



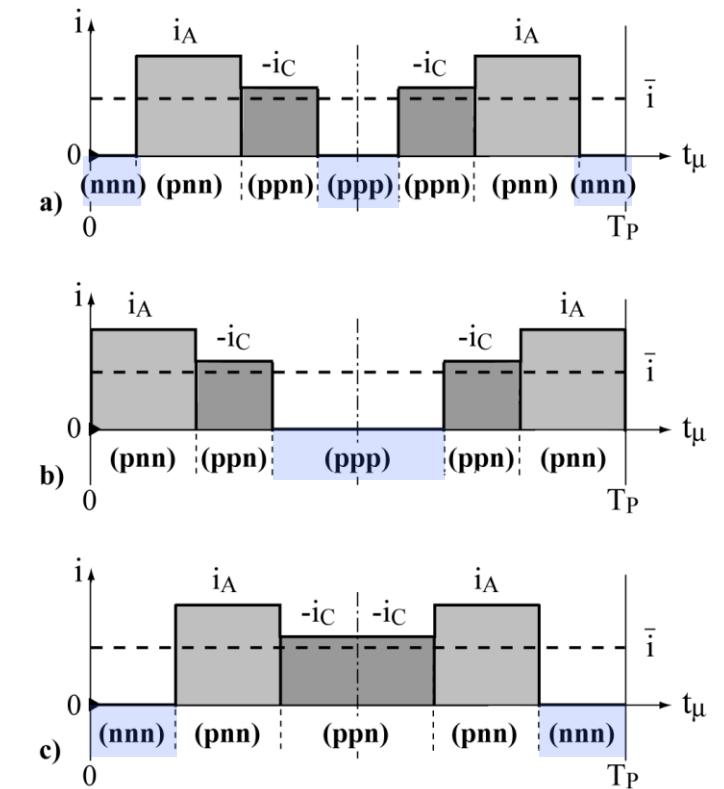
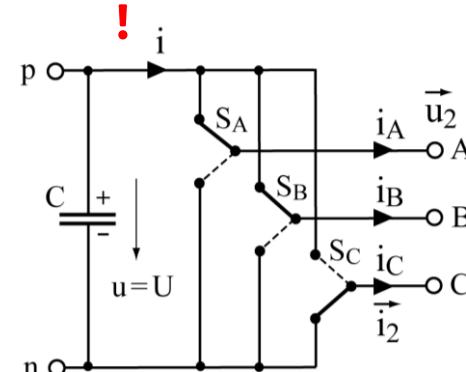
- *Continuous OR Discontinuous Modulation \rightarrow (nnn)-(pnn)-(ppn)-(ppp) OR (nnn)-(pnn)-(ppn)*

VSC Space Vector Modulation (2)

- *Switching Considering Interlock Delay Times*
- $2^3 = 8$ Switching States



$$\bar{i} = I = \frac{3}{4} M_2 \hat{I}_2 \cos \Phi_2$$



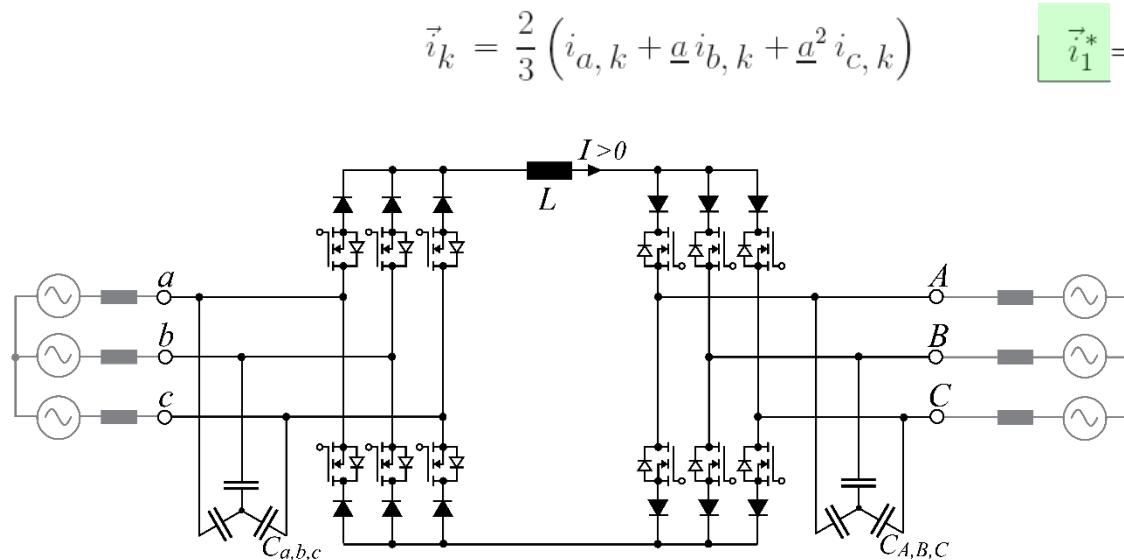
- Continuous OR Discontinuous Modulation \rightarrow (nnn)-(pnn)-(ppn)-(ppp) OR (nnn)-(pnn)-(ppn)



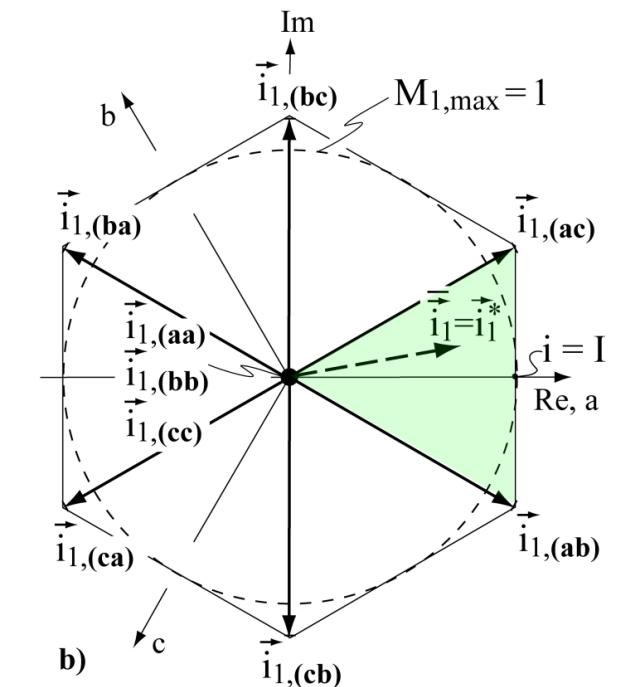
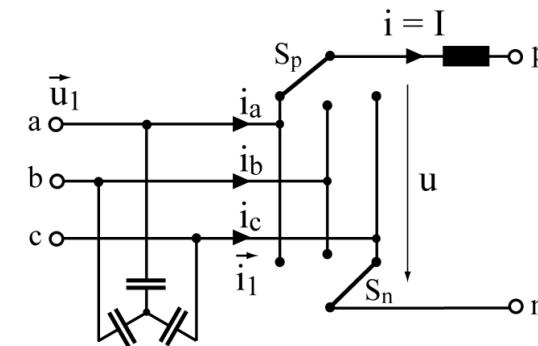
*Current Source
Converter*

CSC Space Vector Modulation (1)

- Overlapping Switching → Natural or Forced Commutation
- $3^2 = 9$ Switching States



$$\vec{i}_1^* = \hat{I}_1^* e^{j\varphi_{\vec{i}_1^*}} = \hat{I}_1^* e^{j(\omega_1 t - \Phi_1^*)}$$

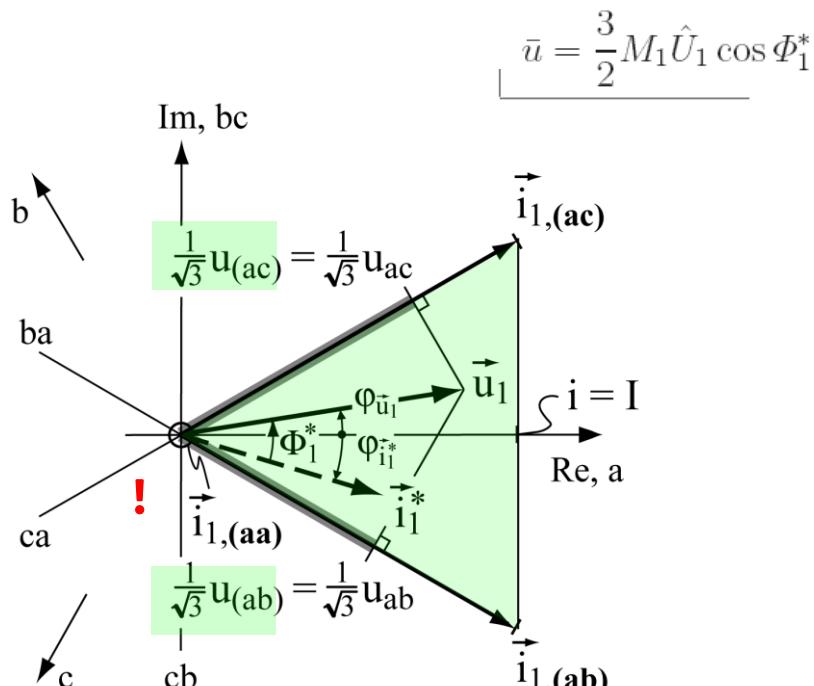
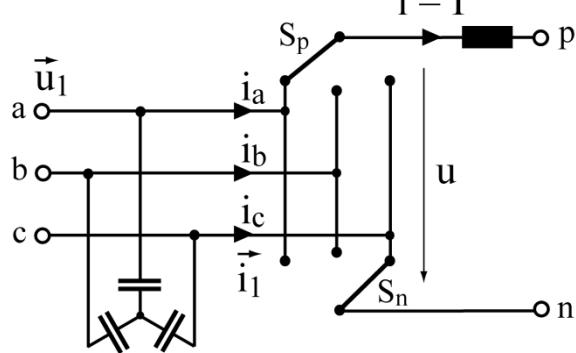


- Shoot-Through Free-Wheeling States (aa), (bb), (cc) → $i_a = i_b = i_c = 0$

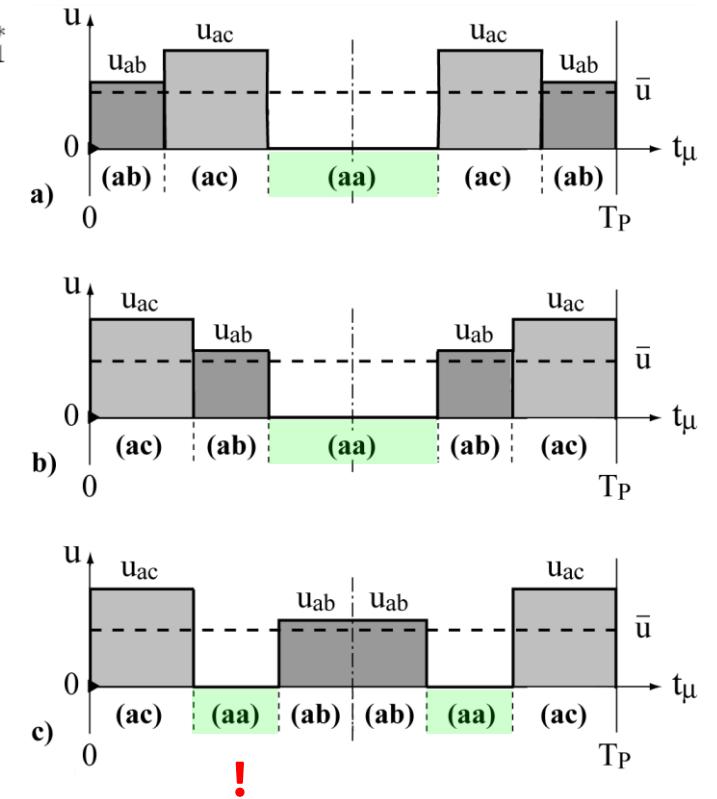


CSC Space Vector Modulation (2)

- Overlapping Switching → Natural or Forced Commutation
- $3^2 = 9$ Switching States



$$\bar{u} = \frac{3}{2} M_1 \hat{U}_1 \cos \Phi_1^*$$



- Shoot-Through Free-Wheeling States (aa), (bb), (cc) $\rightarrow i_a = i_b = i_c = 0$

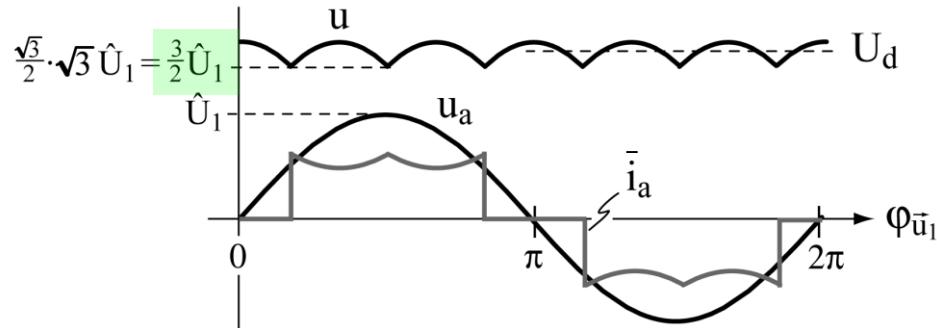


Indirect Matrix Converter

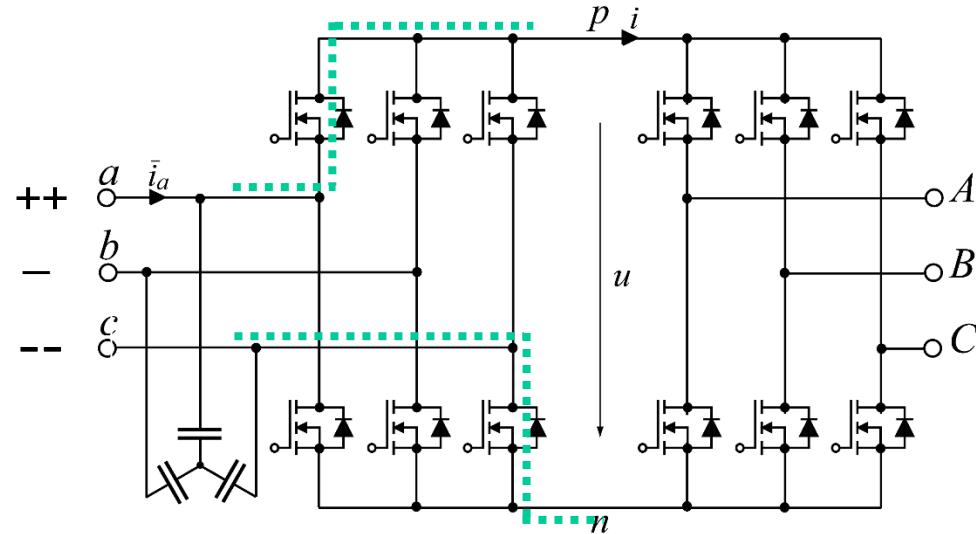
*Topology Derivation
Space Vector Modulation
Waveforms*

Fundamental Frequency Front-End (F^3E)

- Voltage DC-Link AC/AC Converter w/o Energy Storage
- Input Diode Bridge w/ Antiparallel Transistors → Regenerative Braking



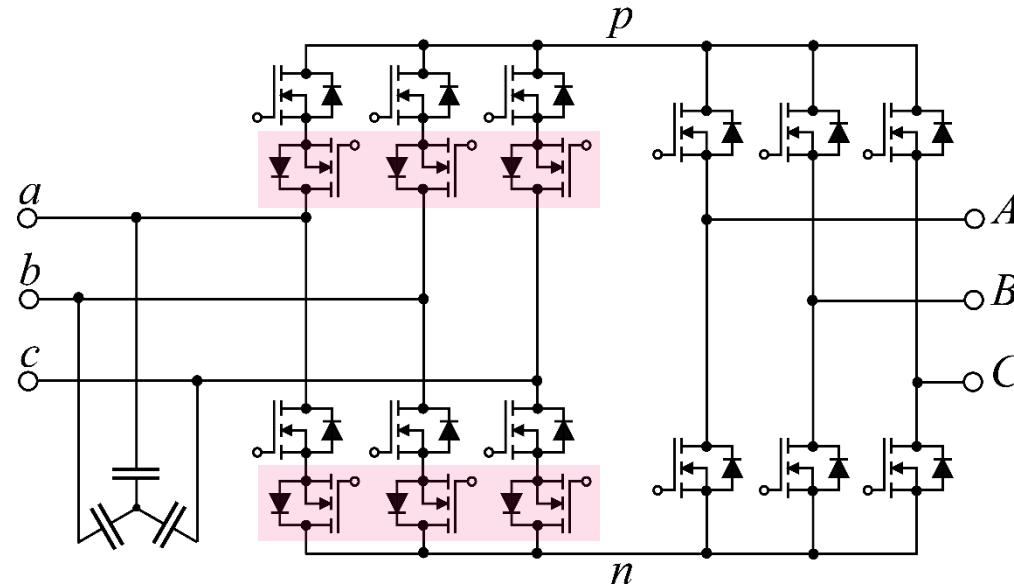
$$\hat{U}_2^* < \frac{\sqrt{3}}{2} \cdot \hat{U}_1 \approx 0.86 \hat{U}_1$$



- Limited Output Voltage Range
- Diodes Determine Switching State of Mains Interface → Block-Shaped Mains Current

Indirect Matrix Converter (IMC)

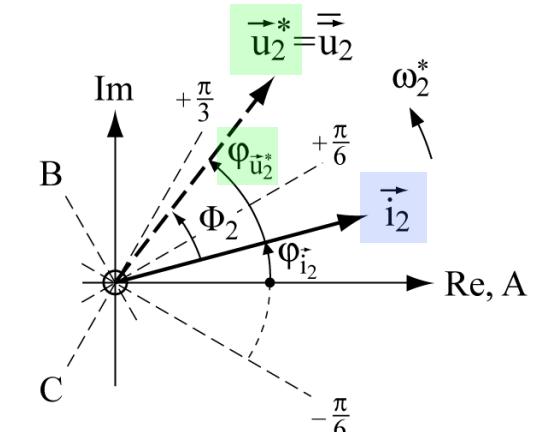
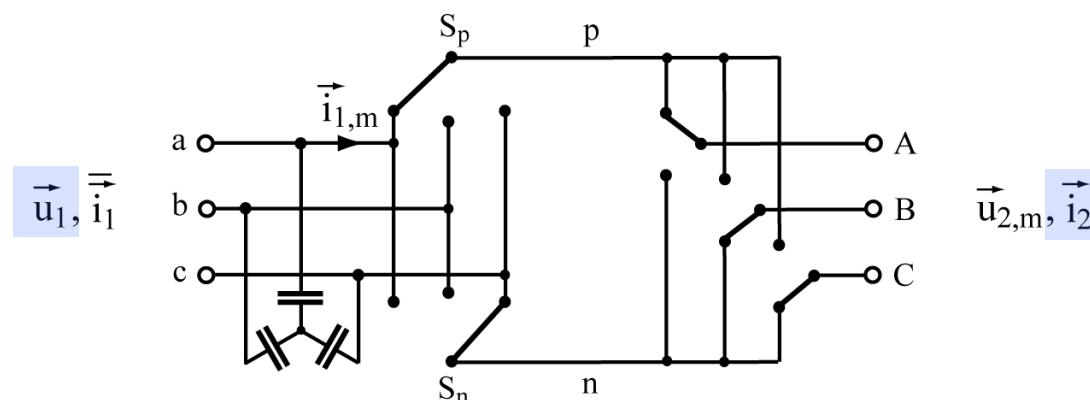
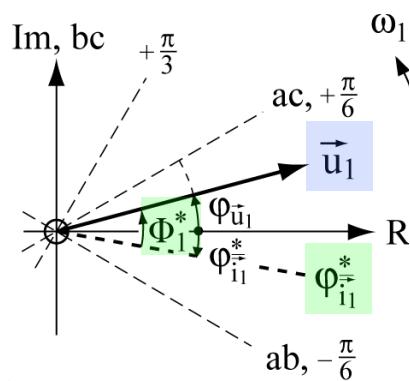
- *Extension of the F³E-Topology → AC Switches / Full Controllability of Mains Interface*
- *Sinusoidal Mains Current*



- *Positive DC-Link Voltage Mandatory !*
- *Coordinated PWM of Input & Output Stage*

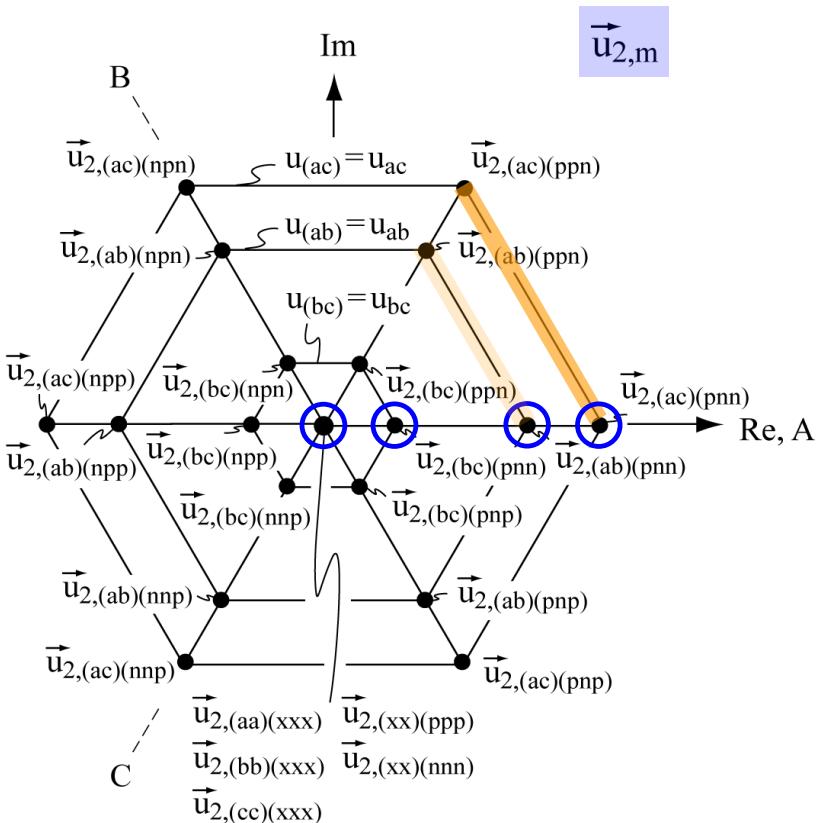
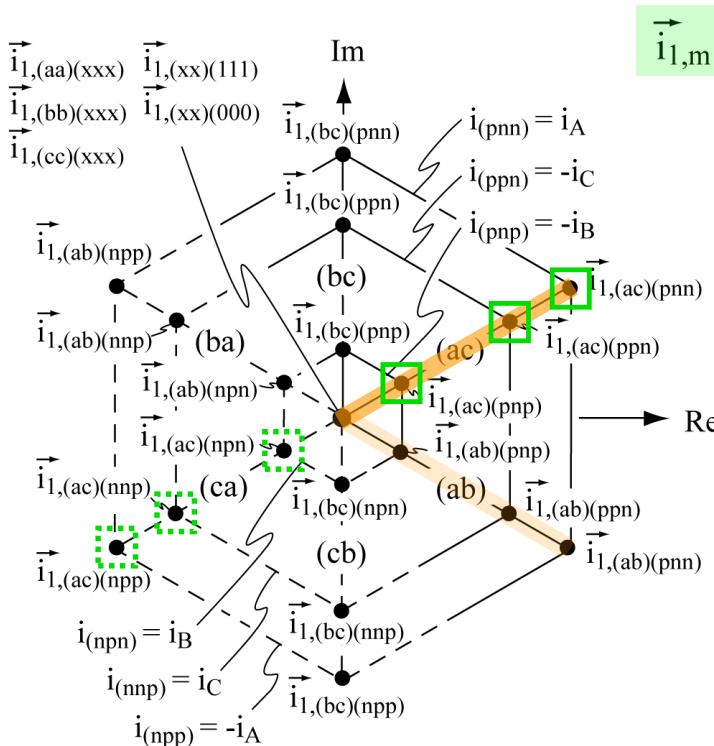
IMC Space Vector Modulation

- Hybrid Voltage DC-Link / Current DC-Link Converter
- Positive DC-Link Voltage Mandatory !



- DC-Link Voltage → Defined by Mains Line-to-Line Voltage Sections
- DC-Link Current → Defined by Load Current Sections

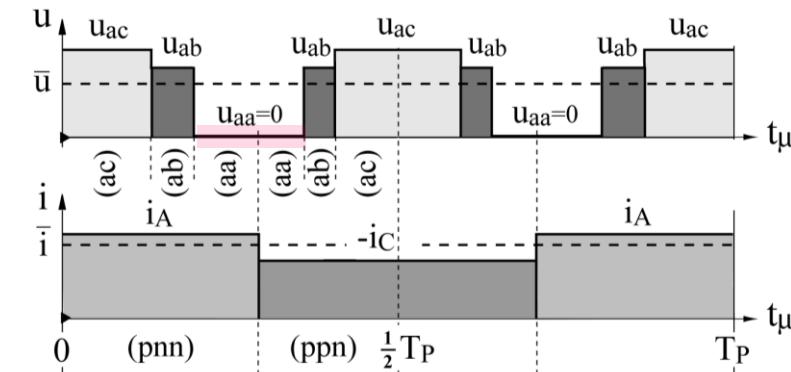
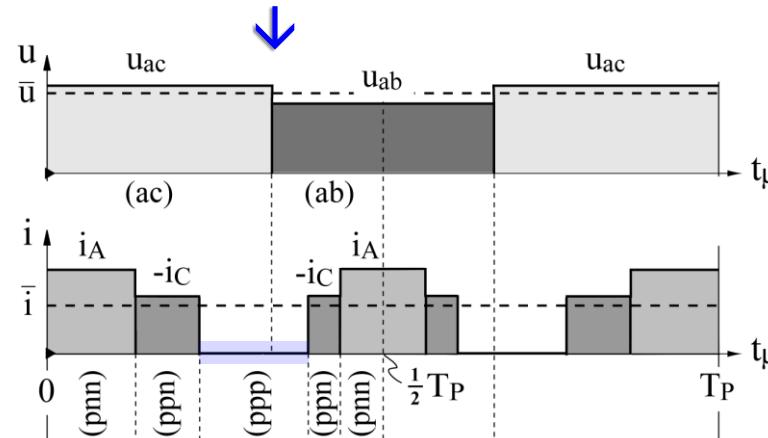
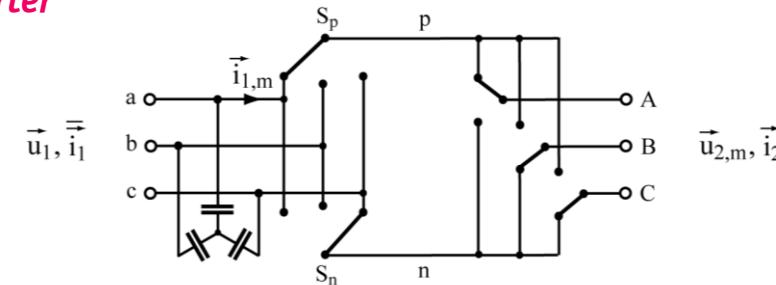
IMC Voltage & Current Space Vectors



- Positive DC-Link Voltage Mandatory!
- Coordinated PWM of Input & Output Stage

IMC Commutation / Modulation

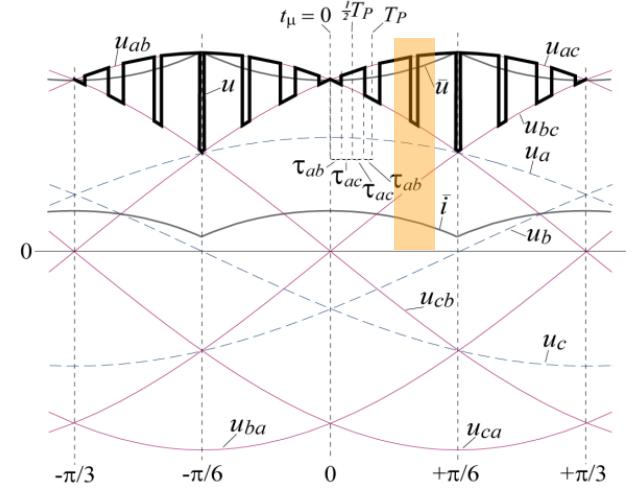
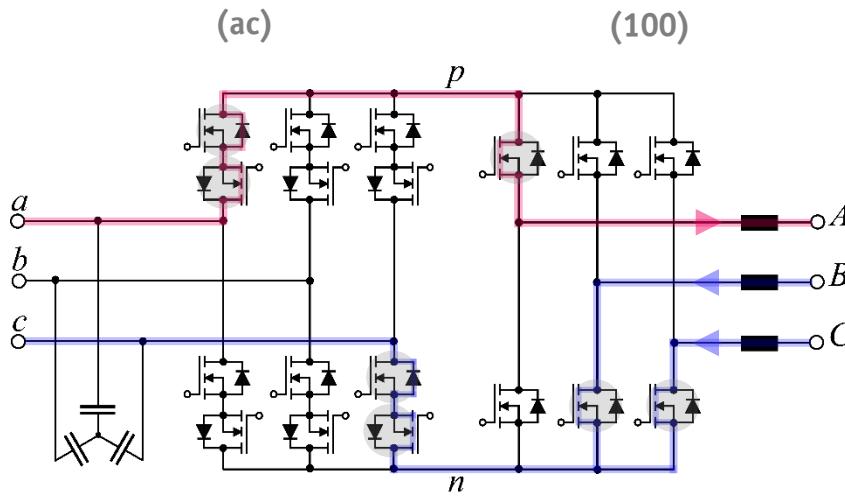
- *Hybrid Voltage DC-Link / Current DC-Link Converter*
- *Positive DC-Link Voltage Mandatory !*



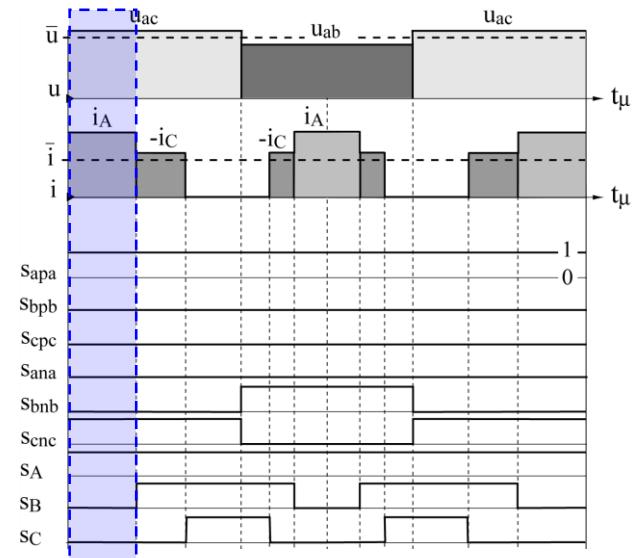
- *Zero Current Commutation of Input Stage*
- *Zero Voltage Commutation of Output Stage*

IMC Modulation (1)

DC-link Voltage $u = u_{ac}$
DC-link Current $i = i_A$



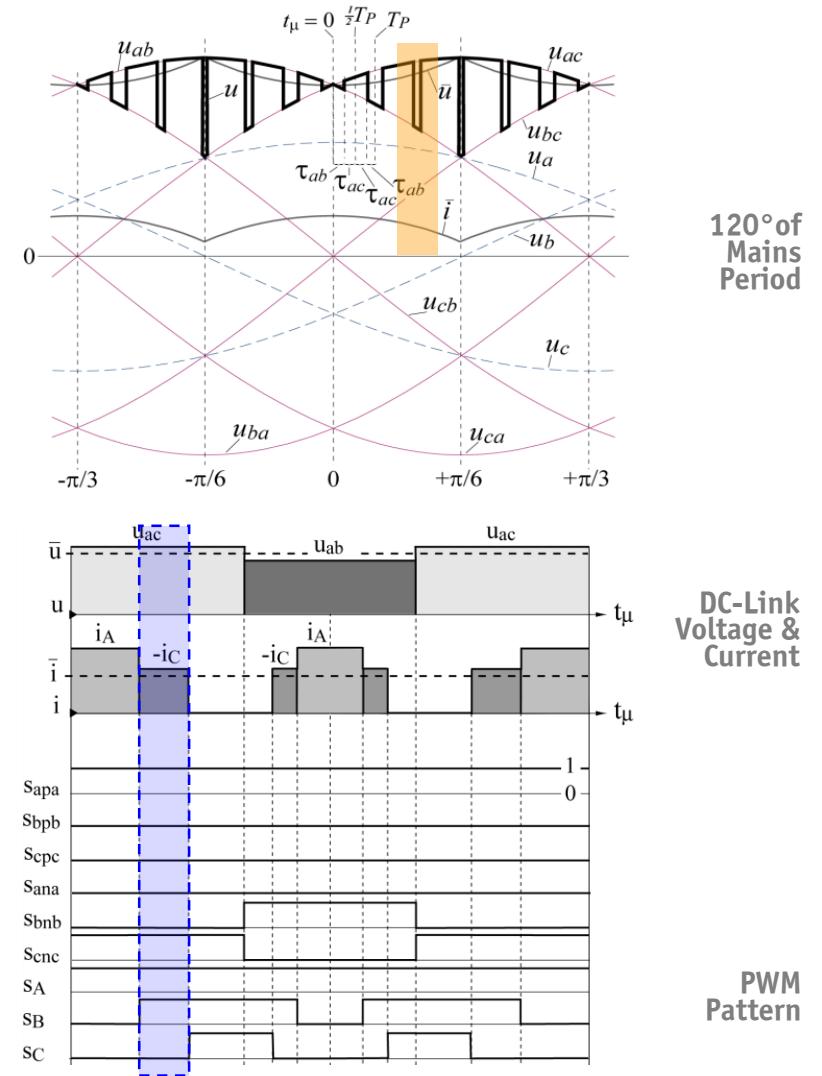
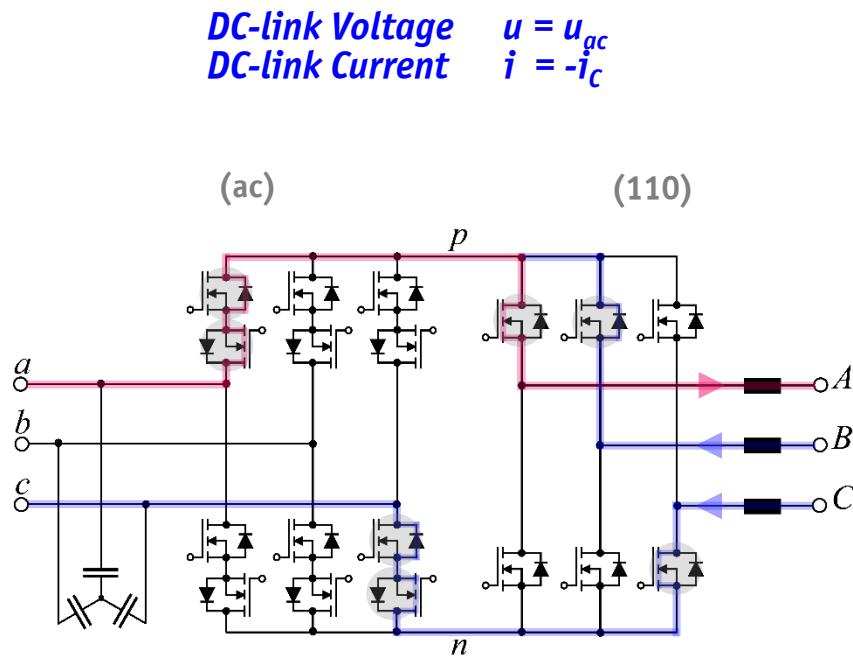
120° of
Mains
Period



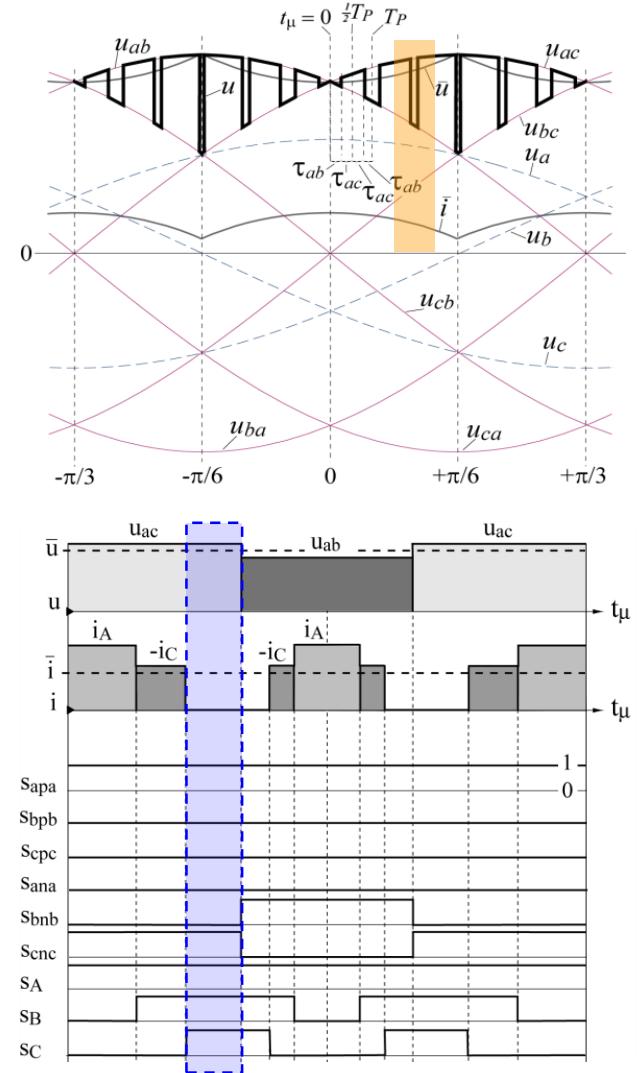
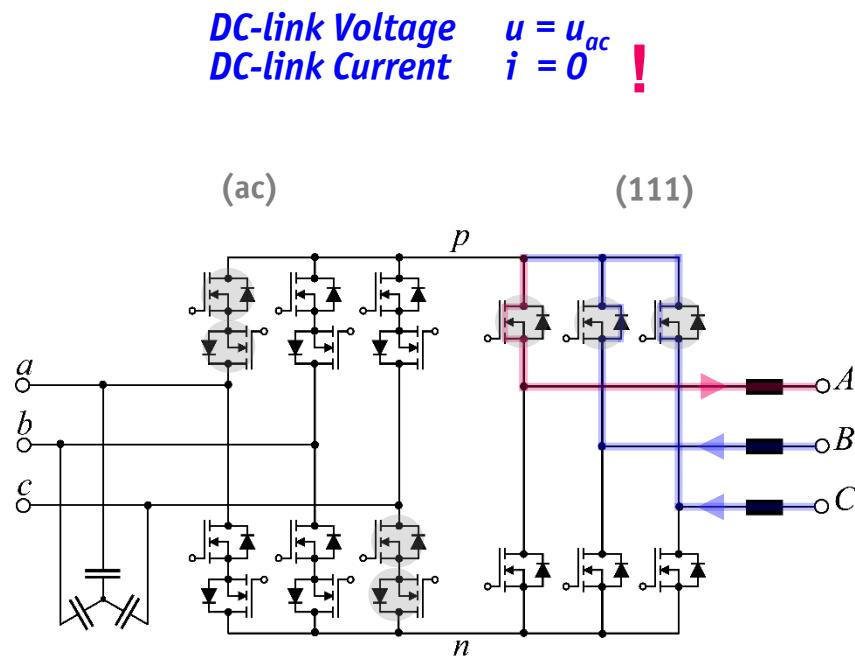
DC link
Voltage &
Current

PWM
Pattern

IMC Modulation (2)



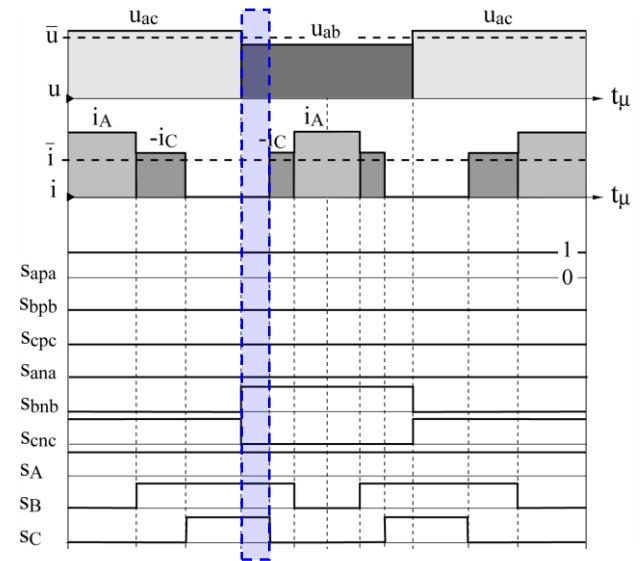
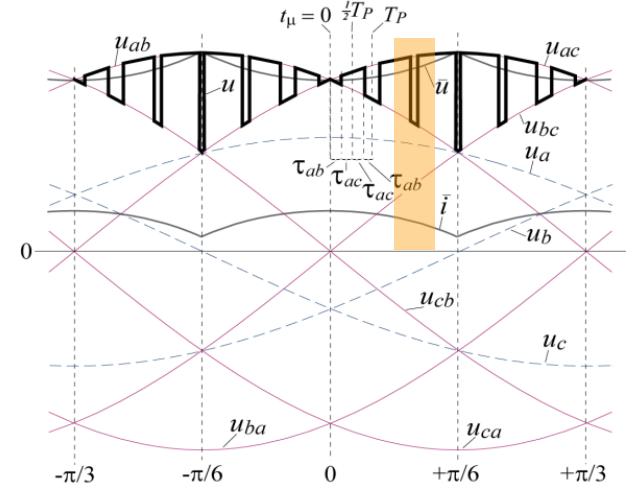
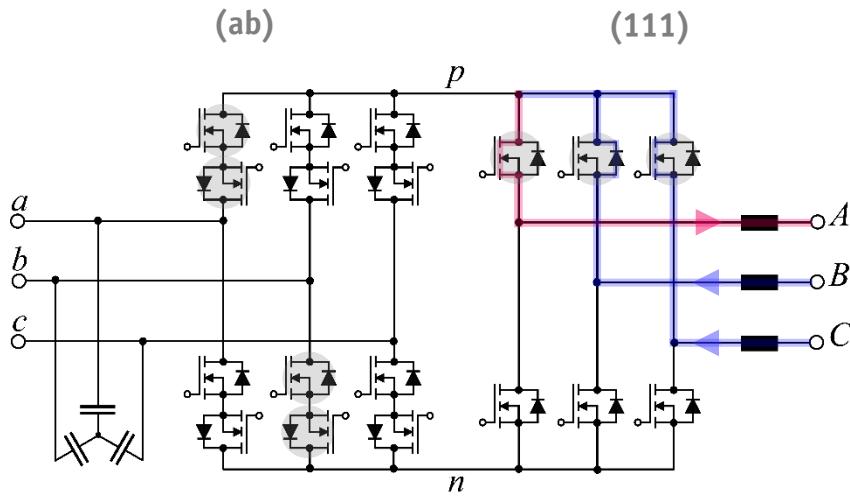
IMC Modulation (3)



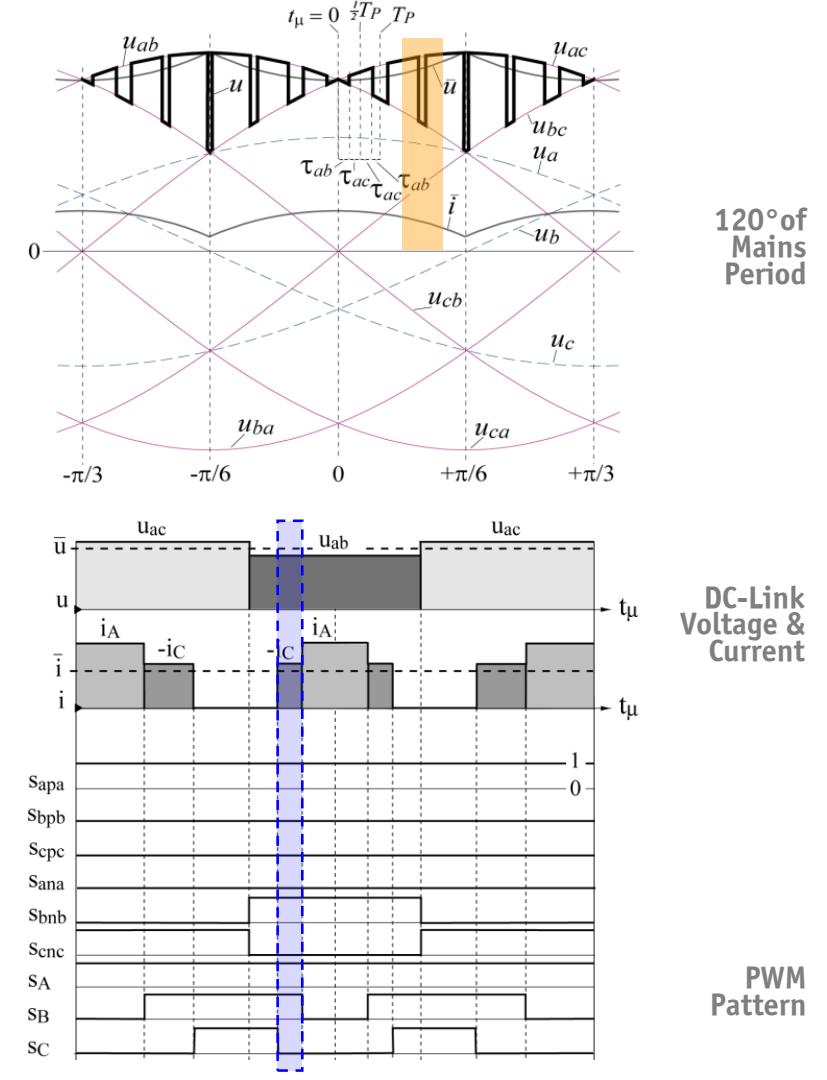
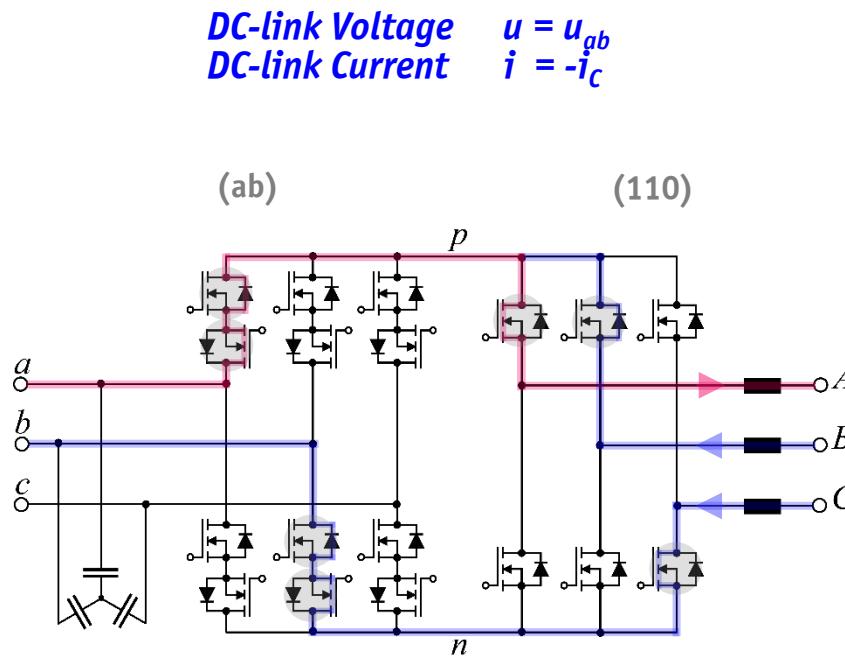
IMC Modulation (3)

DC-link Voltage
DC-link Current

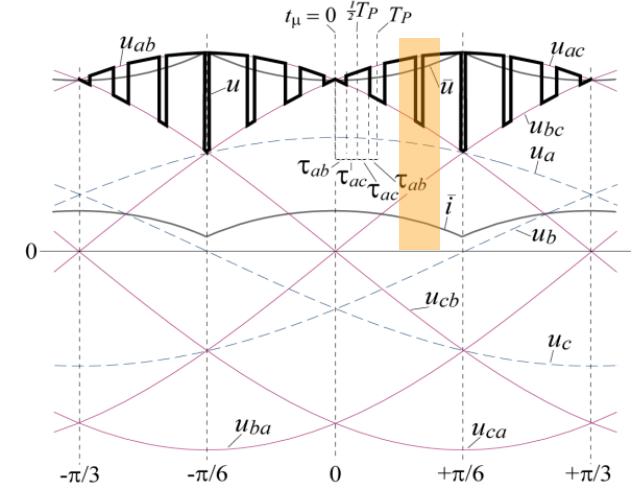
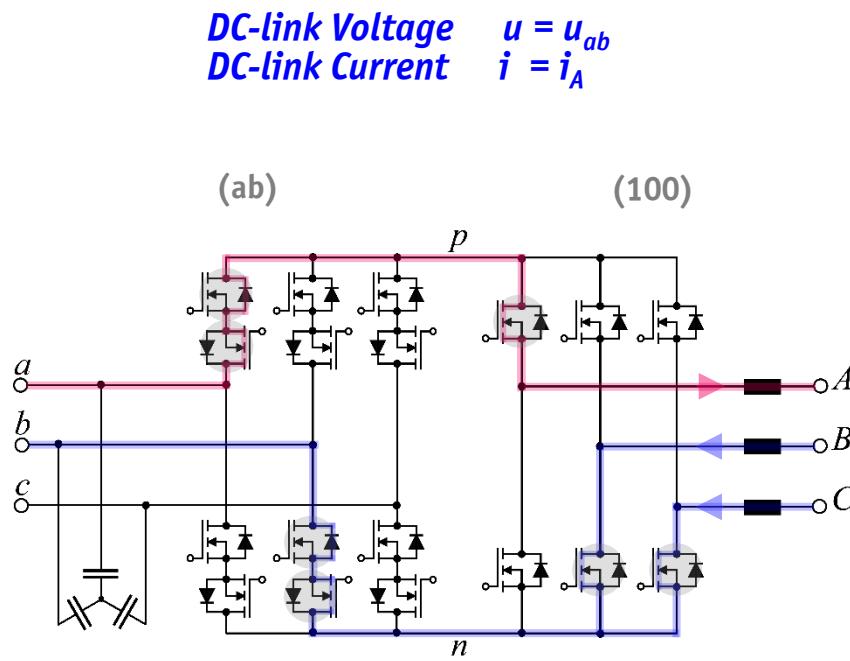
$$u = u_{ab} \quad i = 0 !$$



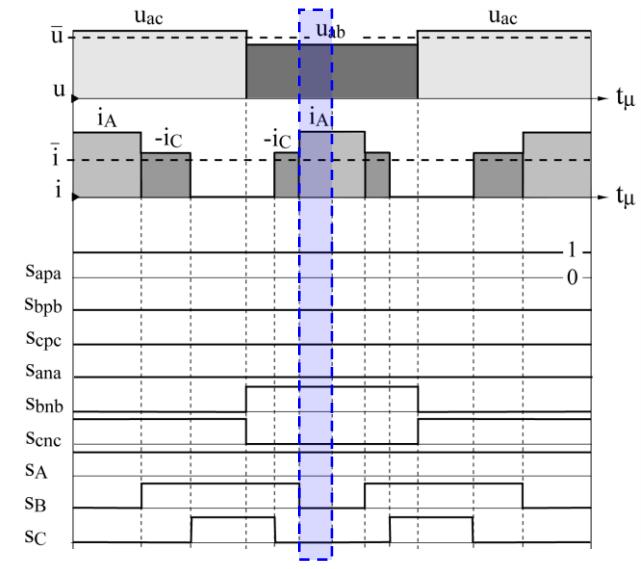
IMC Modulation (5)



IMC Modulation (6)



120° of
Mains Period



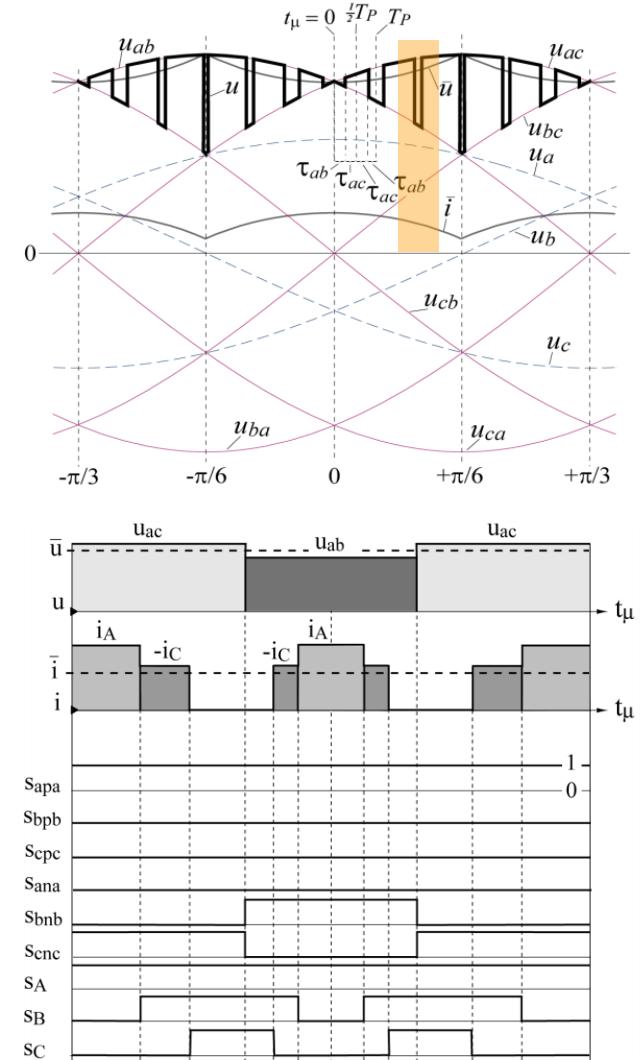
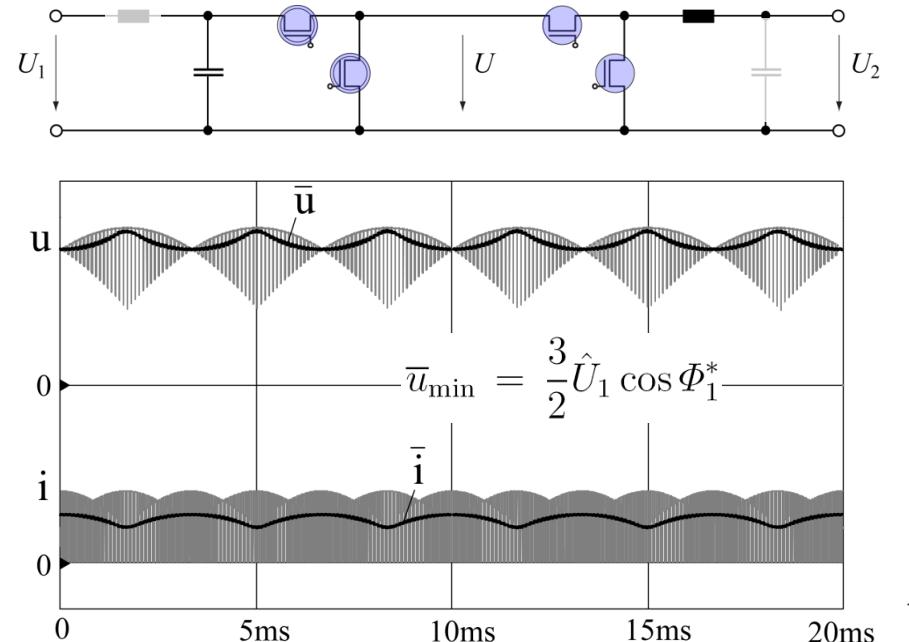
DC-Link
Voltage &
Current

PWM
Pattern

IMC Output Voltage Limit

- IMC — Cascaded Buck-Type Structure
- Input-Stage Output Voltage Reduces with $\cos \Phi_1$
- Analogous to Thyristor AC/DC Converter

$$\hat{U}_{2,\max}^* \leq \frac{\sqrt{3}}{2} \cdot \hat{U}_1 \cos \Phi_1^* !$$



Alternative Modulation Schemes (1)

- *LV vs. HV Modulation*
 - Lower Sw. Losses (40%) & Lower CM Voltage (25%)
 - Slightly Lower Load Current Ripple
 - *Input Voltage Ripple Doubles (!)*
 - Higher Conduction Losses

- *High Output Voltage Modulation (HVM)*

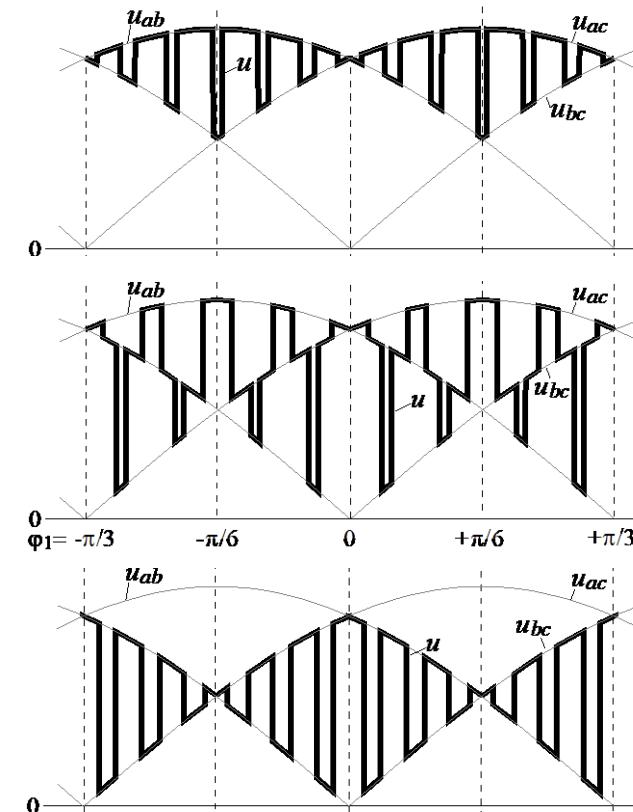
$$\hat{U}_2 = 0 \dots \frac{\sqrt{3}}{2} \cdot \hat{U}_1$$

- *Three-Level Modulation*

$$\hat{U}_2 = \frac{1}{2} \dots \frac{\sqrt{3}}{2} \cdot \hat{U}_1$$

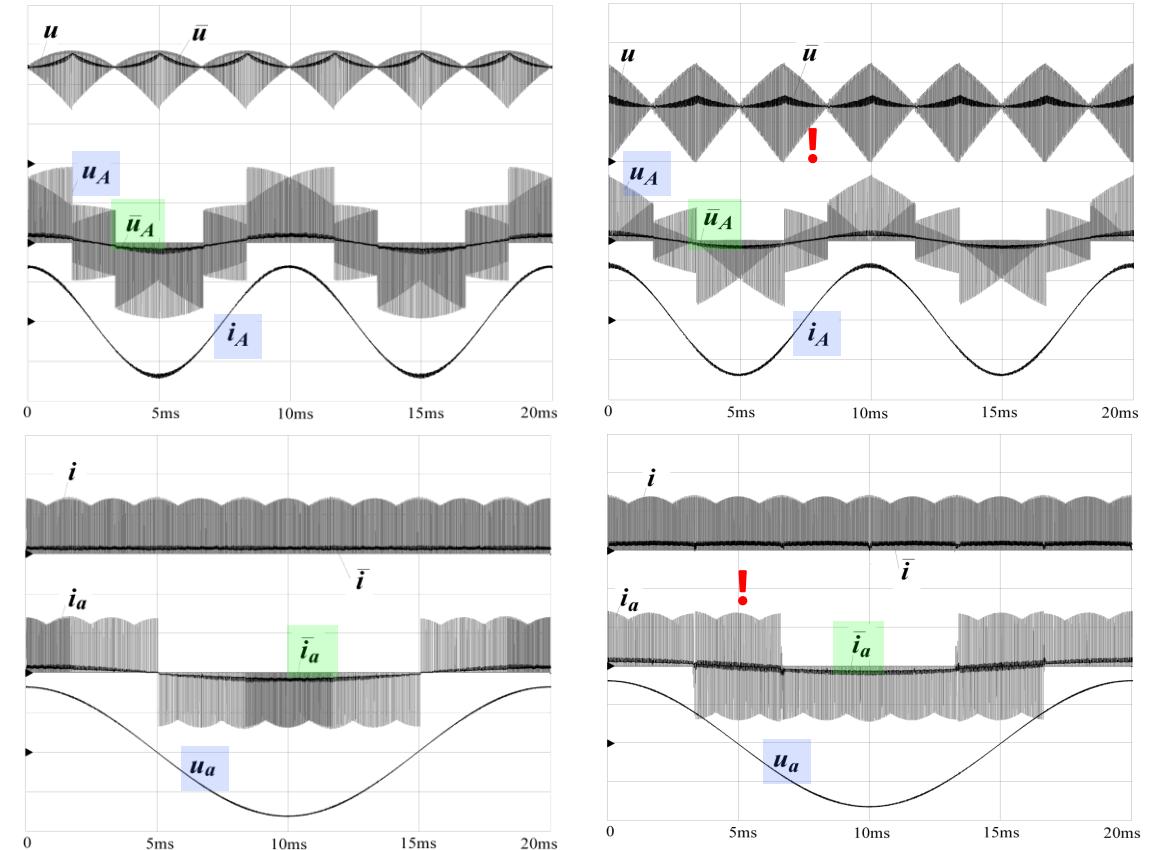
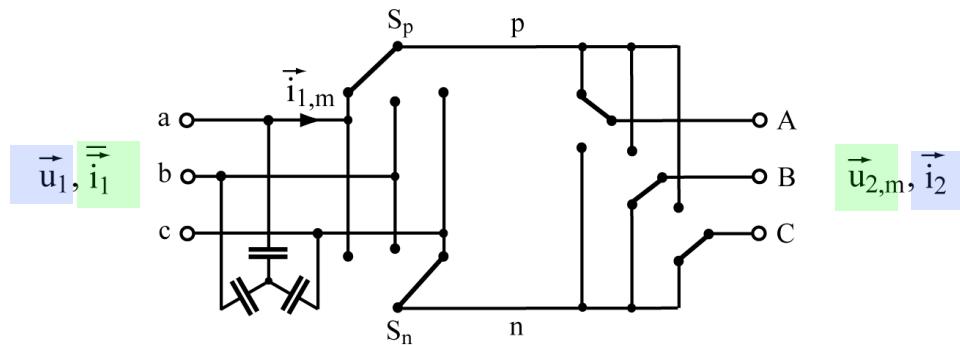
- *Low Output Voltage Modulation (LVM)*

$$\hat{U}_2 = 0 \dots \frac{1}{2} \cdot \hat{U}_1$$



Alternative Modulation Schemes (2)

- HV vs. LV Output Modulation
- Voltage & Current Time Behavior

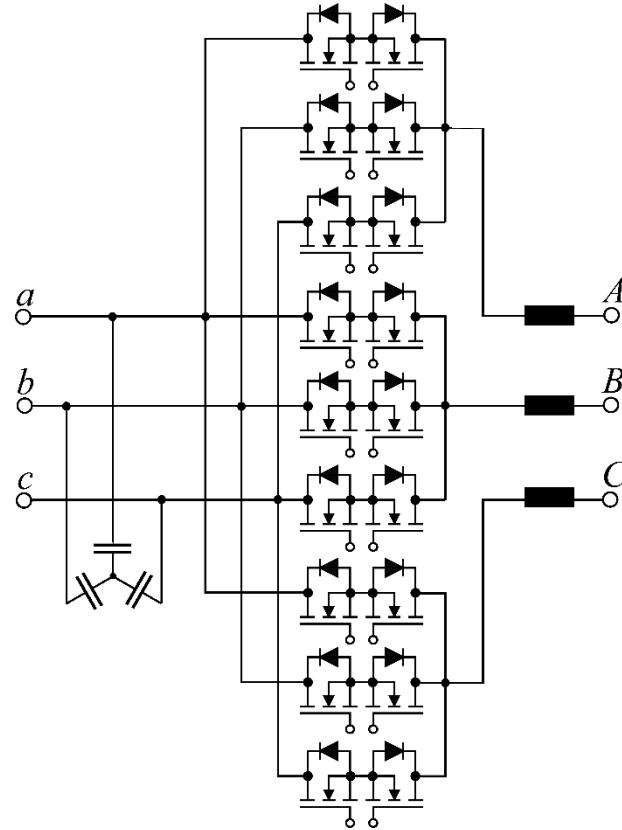




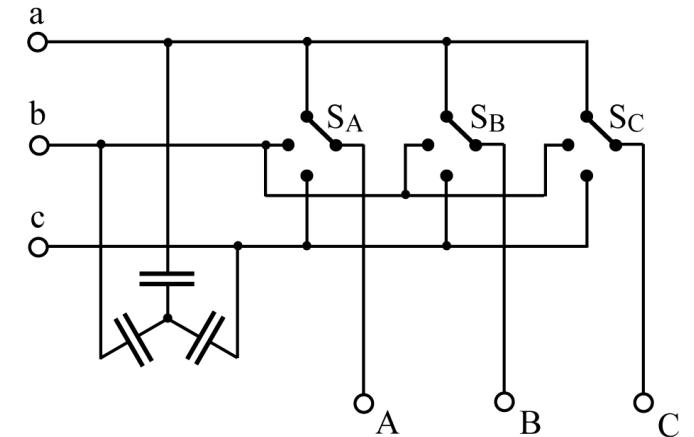
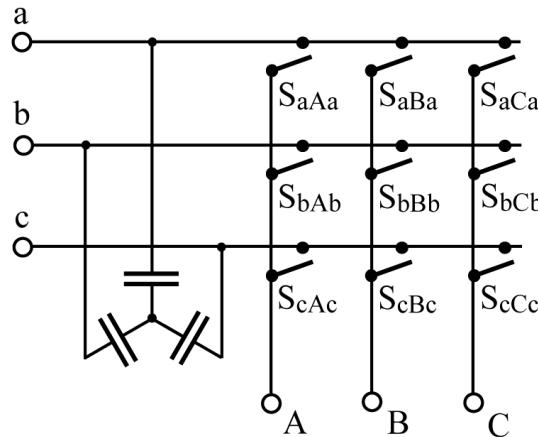
— *Direct Matrix Converter* —

*Multi-Step Commutation
Space Vector Modulation
Waveforms*

Direct Matrix Converter (DMC)



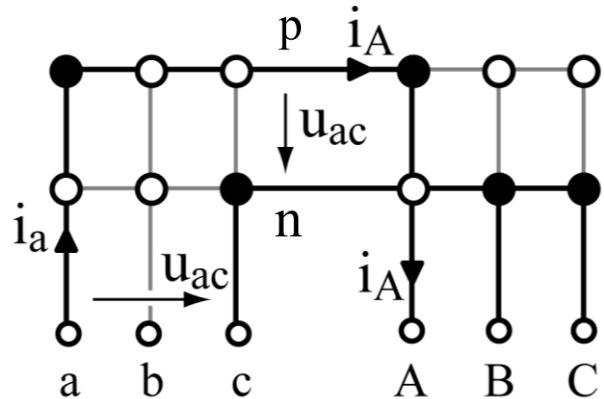
- *Direct AC/AC Conversion → 4-Quadrant (AC) Switches*
- *Quasi Three-Level Output Characteristic*



- *Multi-Step Commutation (!)*
- *Prevent Mains Short Circuit & Interruption of Load Current*

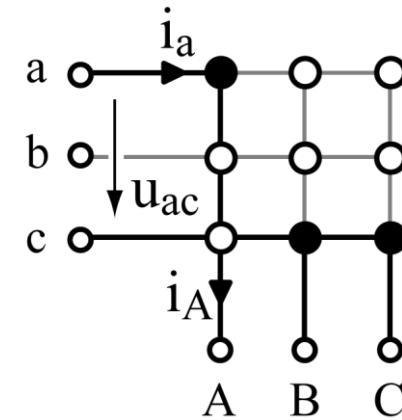
DMC / IMC Relation

- “Indirect” Space Vector Modulation
- Equivalence / Re-Coding of Switching States
- Buck-Type Voltage Conversion



$$\begin{aligned} \dots|_{t_\mu=0} & (ac)(pnn) - (ac)(ppn) - (ac)(ppp) \\ & - (ab)(ppp) - (ab)(ppn) - (ab)(pnn)|_{t_\mu=T_P/2} \\ (ab)(pnn) & - (ab)(ppn) - (ab)(ppp) \\ & - (ac)(ppp) - (ac)(ppn) - (ac)(pnn)|_{t_\mu=T_P} \dots \end{aligned}$$

$$\begin{aligned} \vec{u}_{2,(acc)} &= \vec{u}_{2,(ac)}(pnn) \\ \vec{i}_{1,(acc)} &= \vec{i}_{1,(ac)}(pnn) \end{aligned}$$



$$\begin{aligned} \dots|_{t_\mu=0} & (acc) - (aac) - (aaa) - (aab) - (abb)|_{t_\mu=T_P/2} \\ (abb) - (aab) & - (aaa) - (aaa) - (aac) - (acc)|_{t_\mu=T_P} \dots \end{aligned}$$

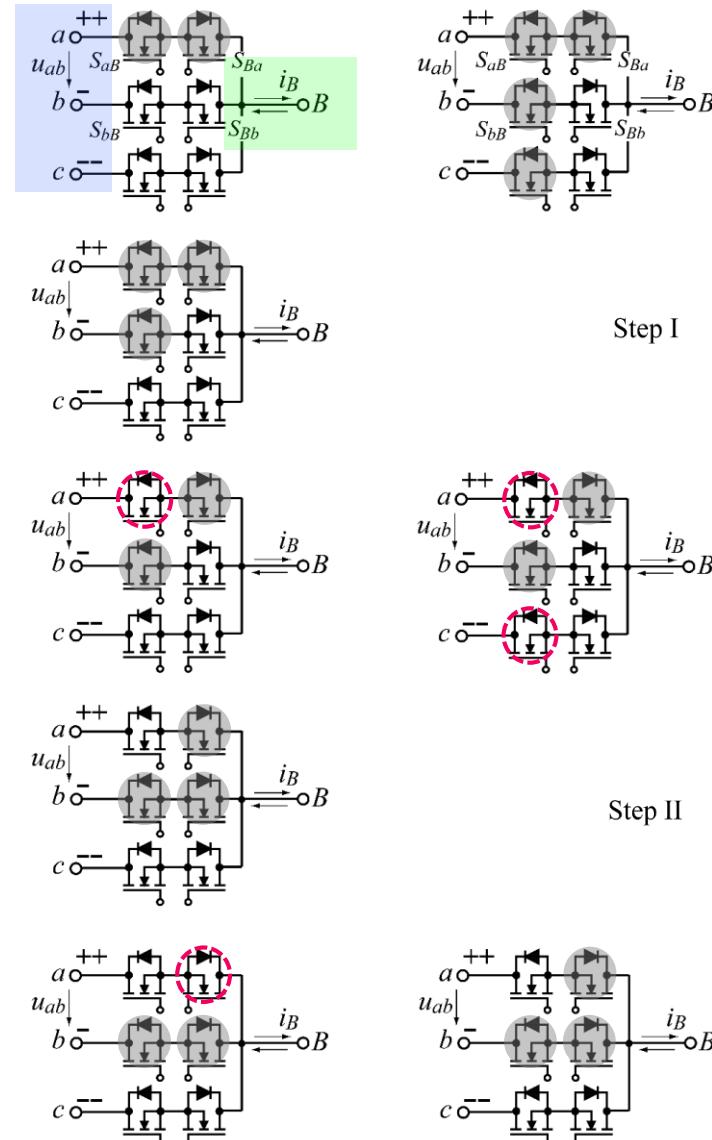
$$\varphi \vec{u}_2^* \in [0, \pi/6]$$

CMC Multi-Step Commutation

Example

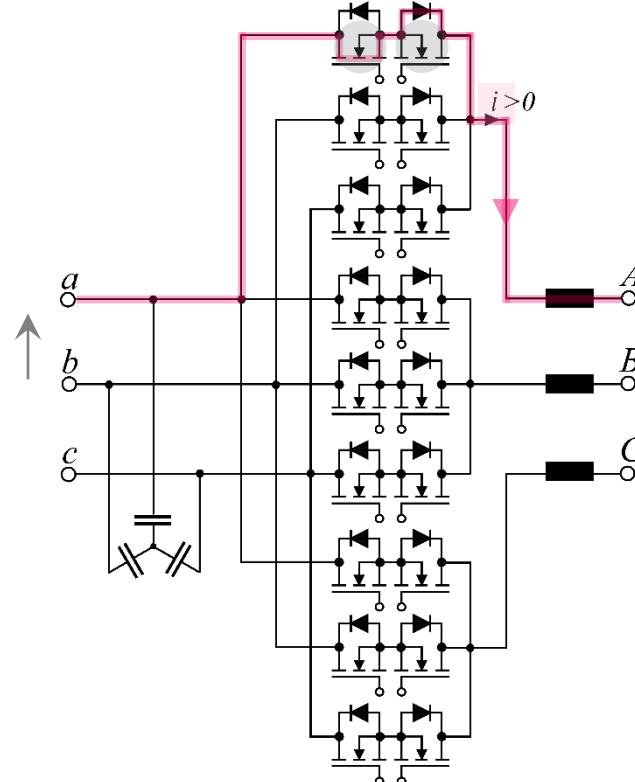
u-Dependent Commutation
 $aB \rightarrow bB$ @
 $u_{ab} > 0$

- *Four-Step Commutation*
- *Two-Step Commutation*



4-Step Commutation of CMC (1)

- Example *i-Dependent Commutation*
 $aA \rightarrow bA @ i > 0$



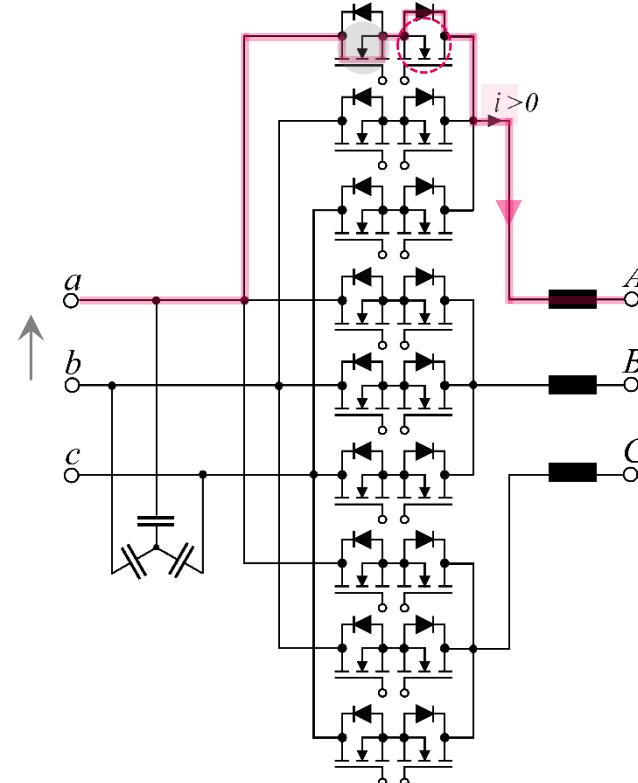
- No Mains Short Circuit
- No Load Current Interruption

Assumption $u_{ab} < 0$

4-Step Commutation of CMC (1)

- Example *i-Dependent Commutation*
 $aA \rightarrow bA @ i > 0$

1st Step: Off



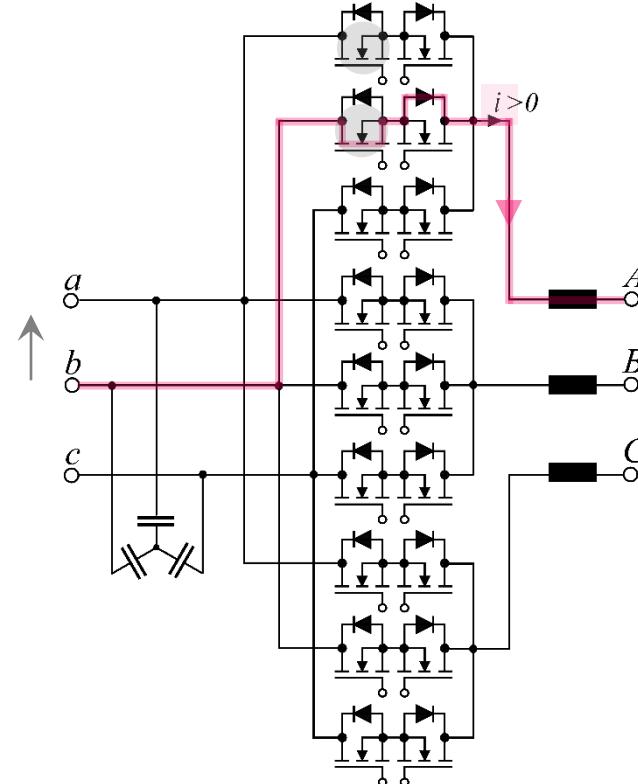
- No Mains Short Circuit
- No Load Current Interruption

Assumption $u_{ab} < 0$

4-Step Commutation of CMC (2)

- Example *i-Dependent Commutation*
 $aA \rightarrow bA @ i > 0$

1st Step: Off
 2nd Step: On



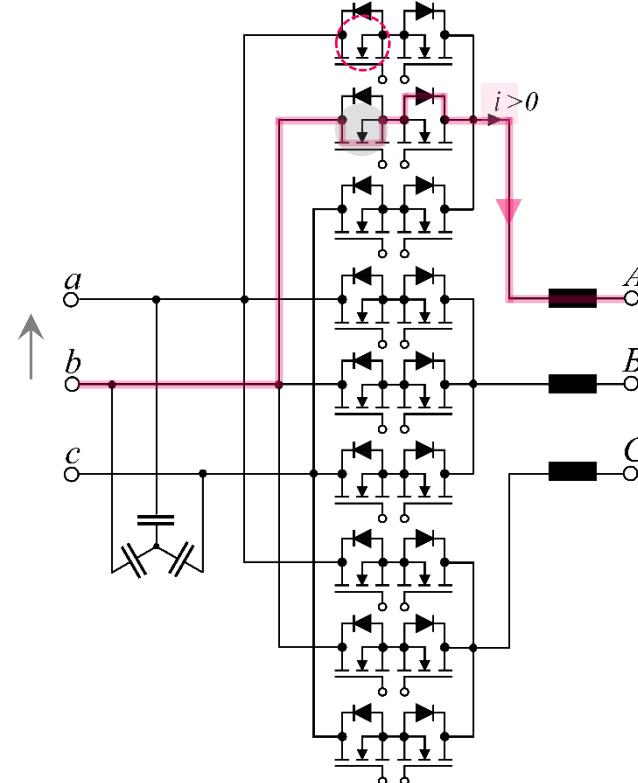
- No Mains Short Circuit
- No Load Current Interruption

Assumption $u_{ab} < 0$

4-Step Commutation of CMC (3)

- Example *i-Dependent Commutation*
 $aA \rightarrow bA @ i > 0$

1st Step: Off
 2nd Step: On
 3rd Step: Off



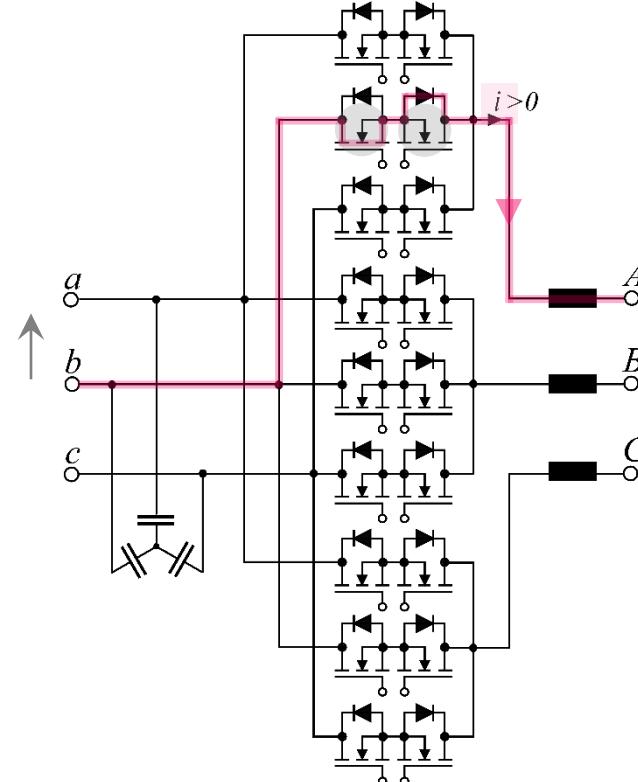
- No Mains Short Circuit
- No Load Current Interruption

Assumption $u_{ab} < 0$

4-Step Commutation of CMC (4)

- Example *i-Dependent Commutation*
 $aA \rightarrow bA @ i > 0$

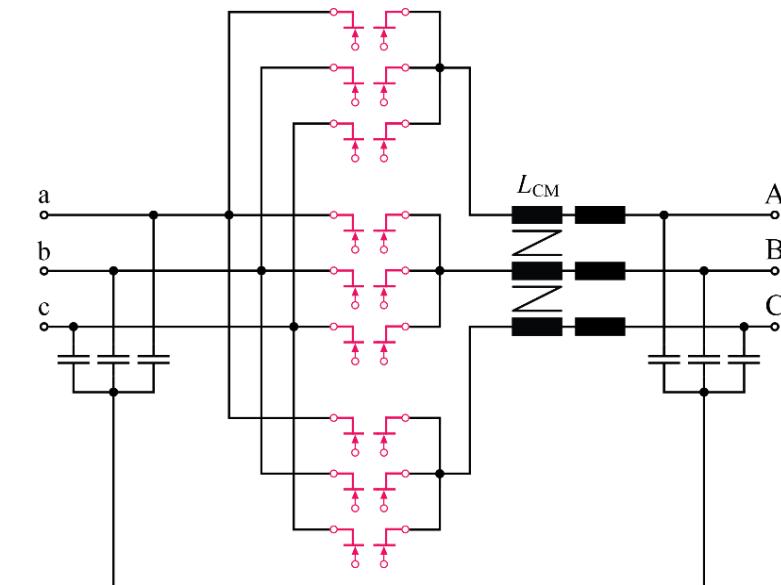
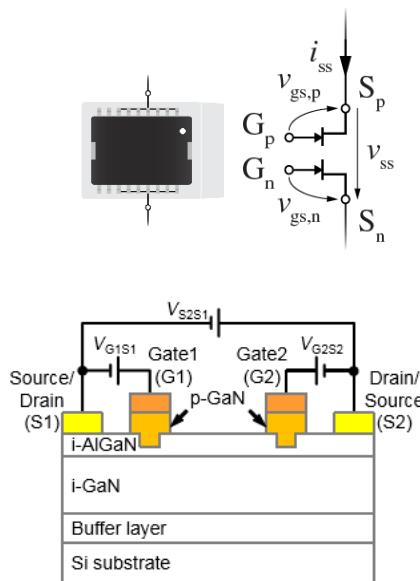
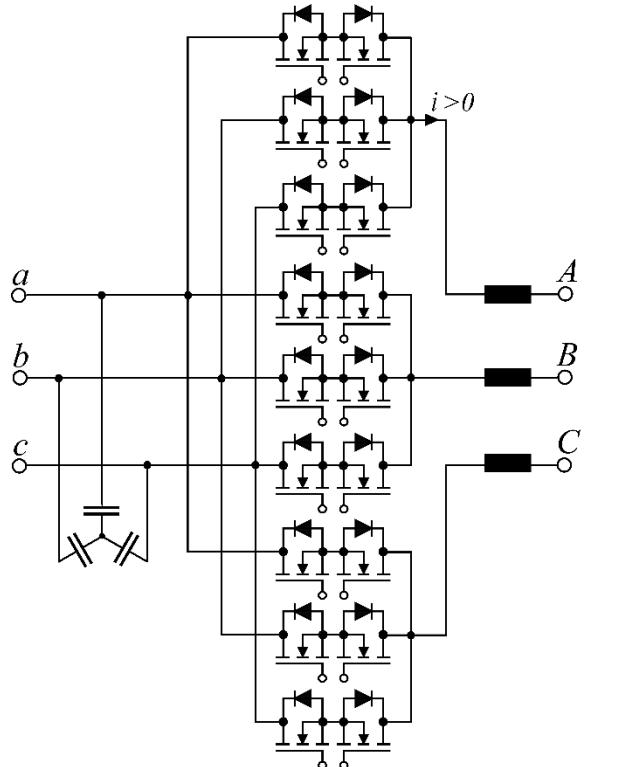
1st Step: Off
 2nd Step: On
 3rd Step: Off
 4th Step: On



- No Mains Short Circuit
- No Load Current Interruption

DMC Employing Monolithic GaN AC-Switches

- *Factor of 4 Saving in Chip Area vs. Discrete Realization*
- *2 Gates → 4-Step Commutation*



- *Full Sinusoidal Output Filter for Preventing Motor Insulation Stress | CM-EMI | Bearing Current*



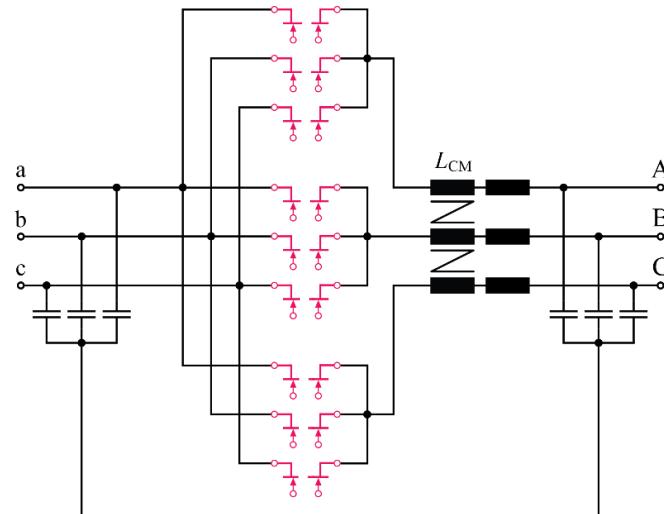
Current Source AC/AC Converter

*Comparative Evaluation vs. CMC
GaN M-BDS Realization*

3-Φ AC/AC Converter Topologies

■ Direct Matrix Converter

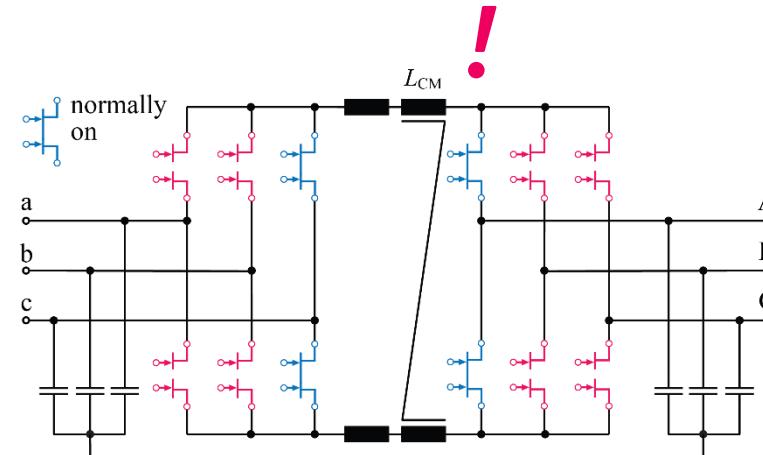
- Application of M-BDSs | 9 Switches
- 4-Step Commutation
- Complex Space Vector Modulation
- Limited to Buck-Operation (!)



- Challenging Overvoltage Protection

■ Current DC-Link Topology

- Application of M-BDSs | 12 Switches
- 4-Step Commutation
- Buck-Boost Functionality (!)
- Low Filter Volume



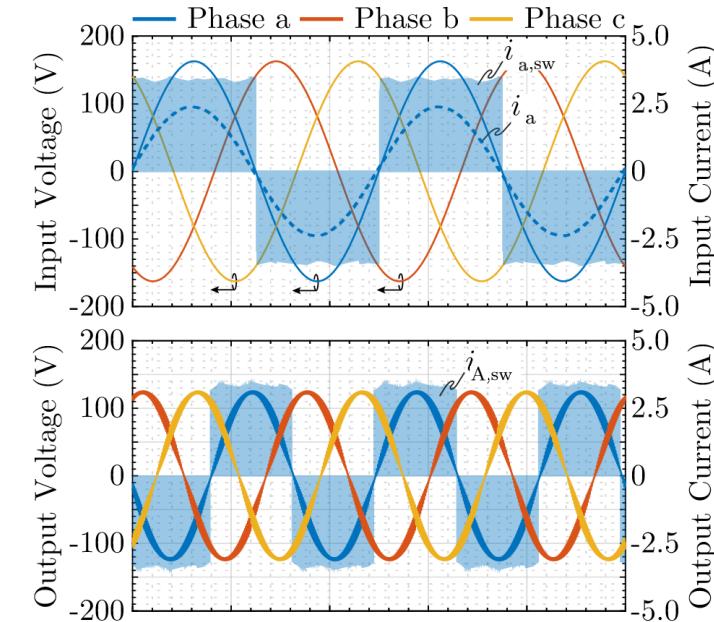
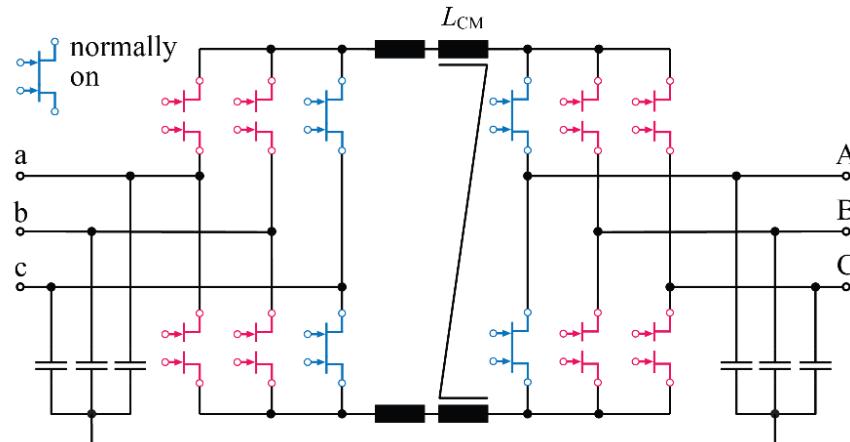
- Challenging Overvoltage Protection

Current DC-Link AC/AC Converter

- Realization w/ **GaN M-BDSs** — 600V/140mΩ
- CM/DM DC-Link Inductor Full-Sinusoidal Filtering
- Protection w/ **Normally-On Switches**

— Compressor Drive for More Electric Aircraft

$$\begin{aligned}\rho &= 10 \text{ kW/dm}^3 \\ \eta &= 98\% \\ f_s &= 72 \text{ kHz}\end{aligned}$$

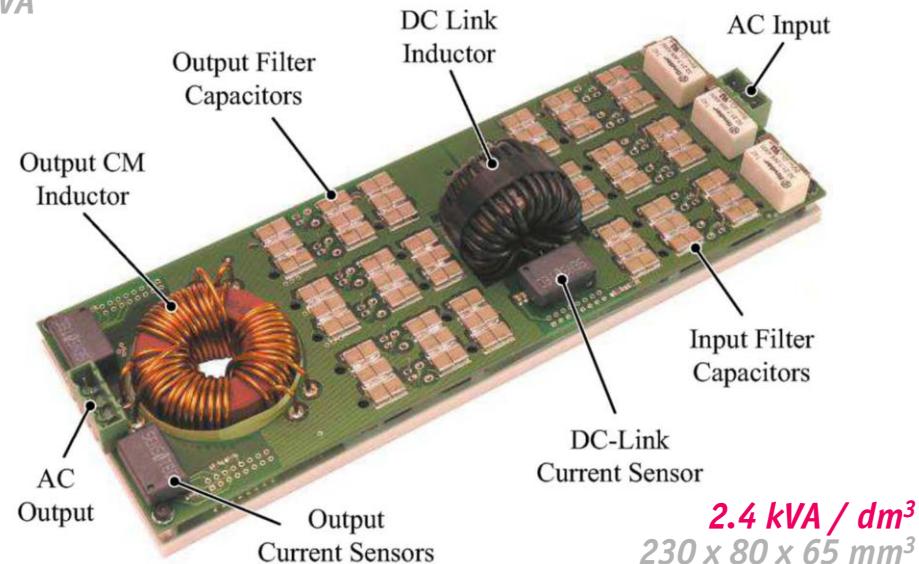


- Advanced Modulation Schemes → Significant Performance Improvement

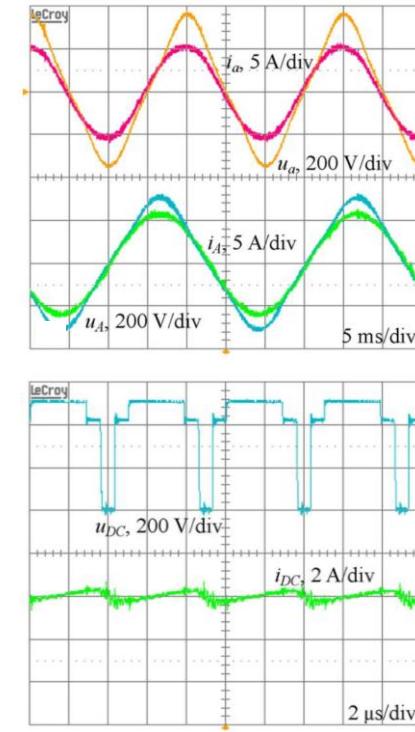
200kHz SiC Current DC-Link AC/AC Converter

- *SiC J-FETs — Normally-On T0-220 1200V/6A — Built in 2008 (!)*
- *1200V/10A SiC Schottky Series Diodes*
- *X7R Ceramic Filter Capacitors*

Input $400V_{rms}$ Line-to-Line
Output $0...300Hz$
Rated Power 2.5 kVA



- *Low Volume DC-Link Inductor (320uH)*



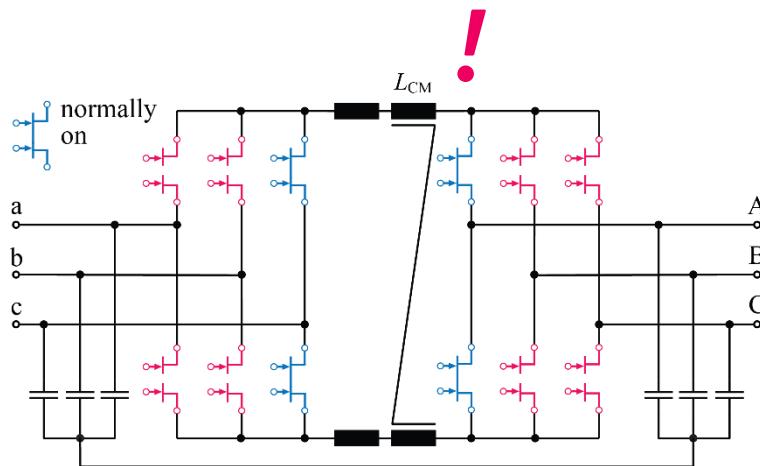


— Conclusion —

3-Φ AC/AC Converter Topologies

■ Current DC-Link Topology

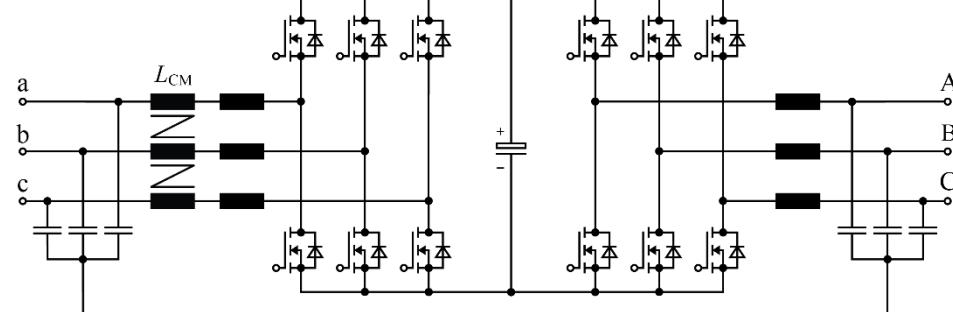
- Application of M-BDSs
- Complex 4-Step Commutation
- Low Filter Volume



- Challenging Overvoltage Protection
- Limited Control Dynamics

■ Voltage DC-Link Topology

- Standard Bridge-Legs
- Low-Complexity Commutation
- Defined Semiconductor Voltage Stress
- Facilitates DC-Link Power Buffer



- High Input / Output Filter Volume



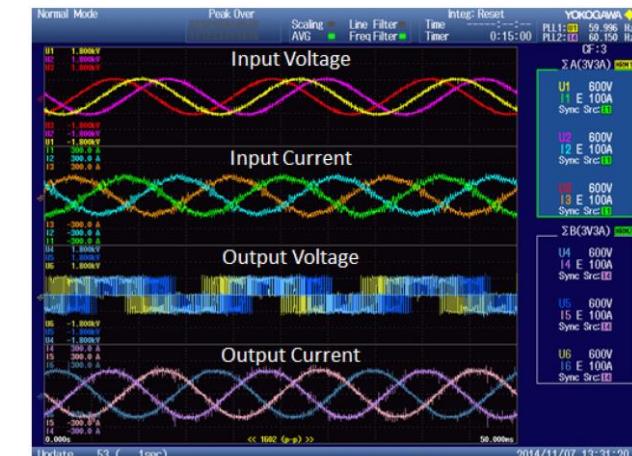
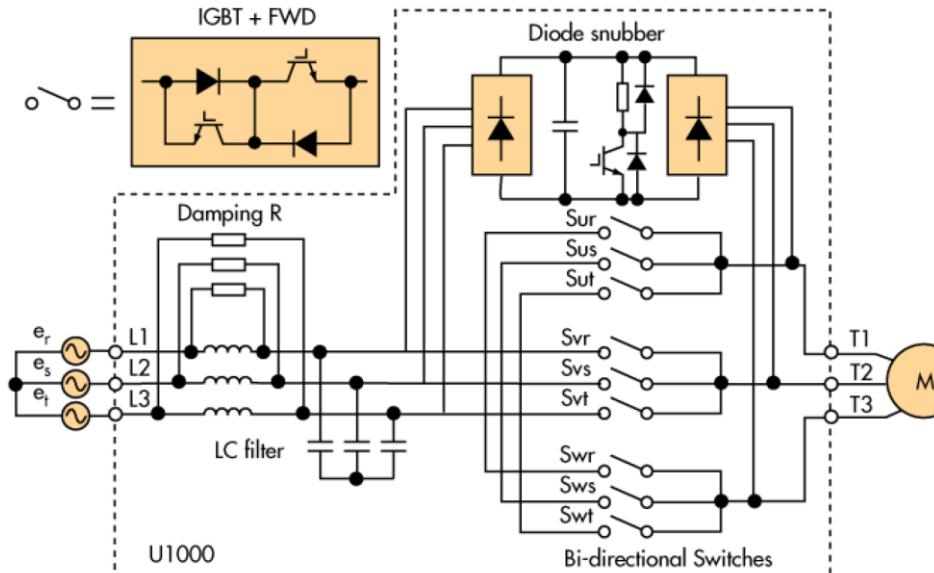
Thank you!





Industry Application of 3-Φ Matrix Converter

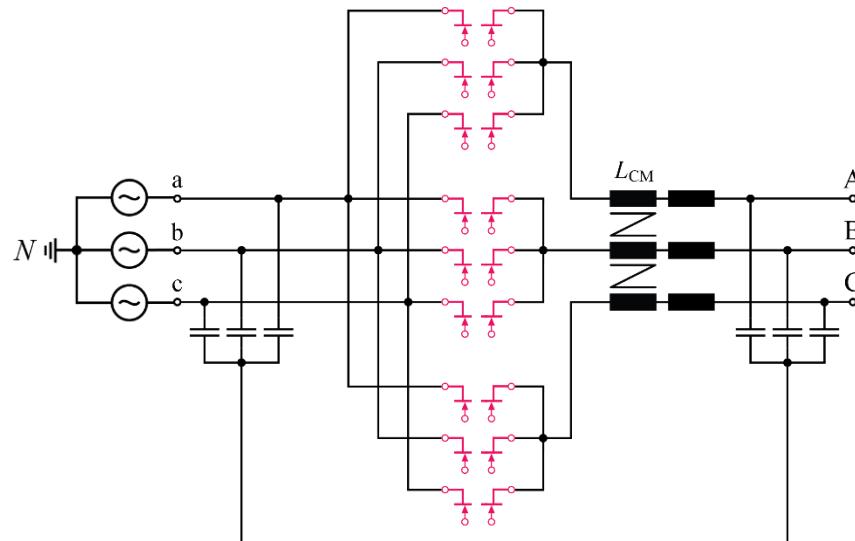
- *Direct Matrix Converter Topology*
- *Fully Regenerative → e.g. Downhill Conveyor etc.*
- *Higher Power Density Compared to Voltage DC Link System / No Front-End Boost Inductors*
- *Quasi Three-Level Output Characteristic*
- *No-Switching Eco Operation for $f_2=f_{\text{Mains}}$*
- *Close to Unity Power Factor*



- *Challenging Overvoltage Protection*
- *Limited Output Voltage Range*

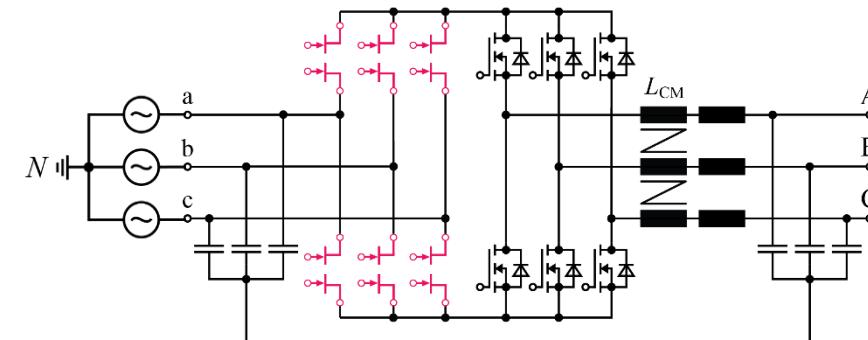
3-Φ AC/AC Matrix Converter

- *Direct Matrix Converter (CMC)*
- 4-Step Commutation
- Exclusive Use of **GaN M-BDSs**



— Thermally Critical @ $f_{out} \approx f_{in}$

- *Indirect Matrix Converter (IMC)*
- CSI GaN M-BDS AC/DC Front-End
- ZCS Commutation of CSI Stage @ $i_{DC}=0$
- No 4-Step Commutation



- Higher # of Switches Compared to CMC
- Lower Cond. Losses @ Low Output Voltage
- Thermally Critical @ $f_{out} \rightarrow 0$